Digital Design

Chapter 4:

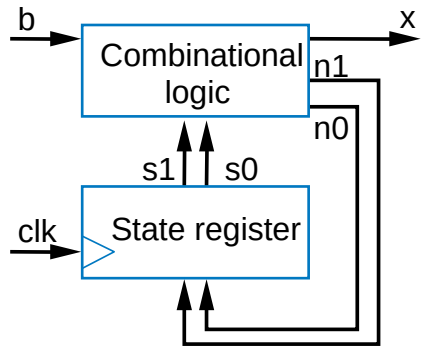
Datapath Components

Introduction

- Chapters 2 & 3: Introduced increasingly complex digital building blocks
  - Gates, multiplexors, decoders, basic registers, and controllers
- Controllers good for systems with control inputs/outputs
  - ***Control*** input: Single bit (or just a few), representing environment event or state
    - e.g., 1 bit representing button pressed
  - ***Data*** input: Multiple bits collectively representing single entity
    - e.g., 7 bits representing temperature in binary
- Need building blocks for data
  - ***Datapath components***, aka register-transfer-level (RTL) components, store/transform data
    - Put datapath components together to form a ***datapath***
- This chapter introduces numerous datapath components, and simple datapaths
  - Next chapter will combine controllers and datapaths into “processors”

Registers

- Can store data, very common in datapaths
- Basic register of Ch 3: Loaded every cycle
  - Useful for implementing FSM -- stores encoded state
  - For other uses, may want to load only on certain cycles



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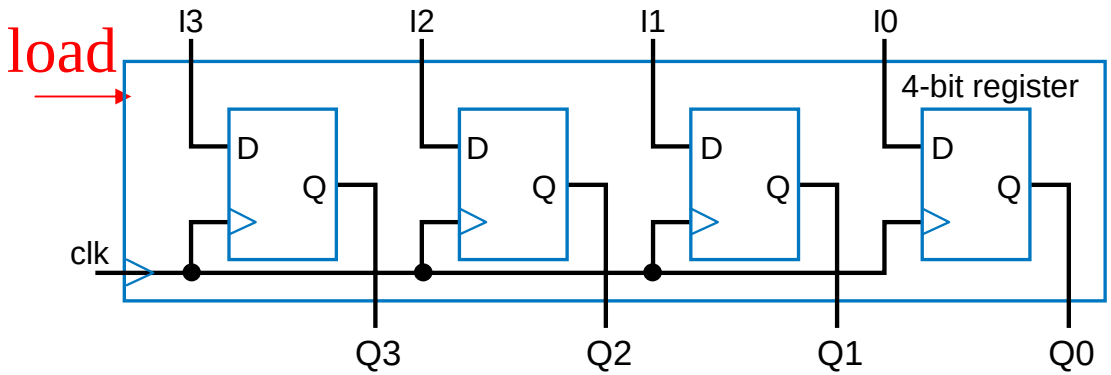
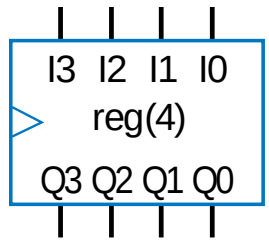


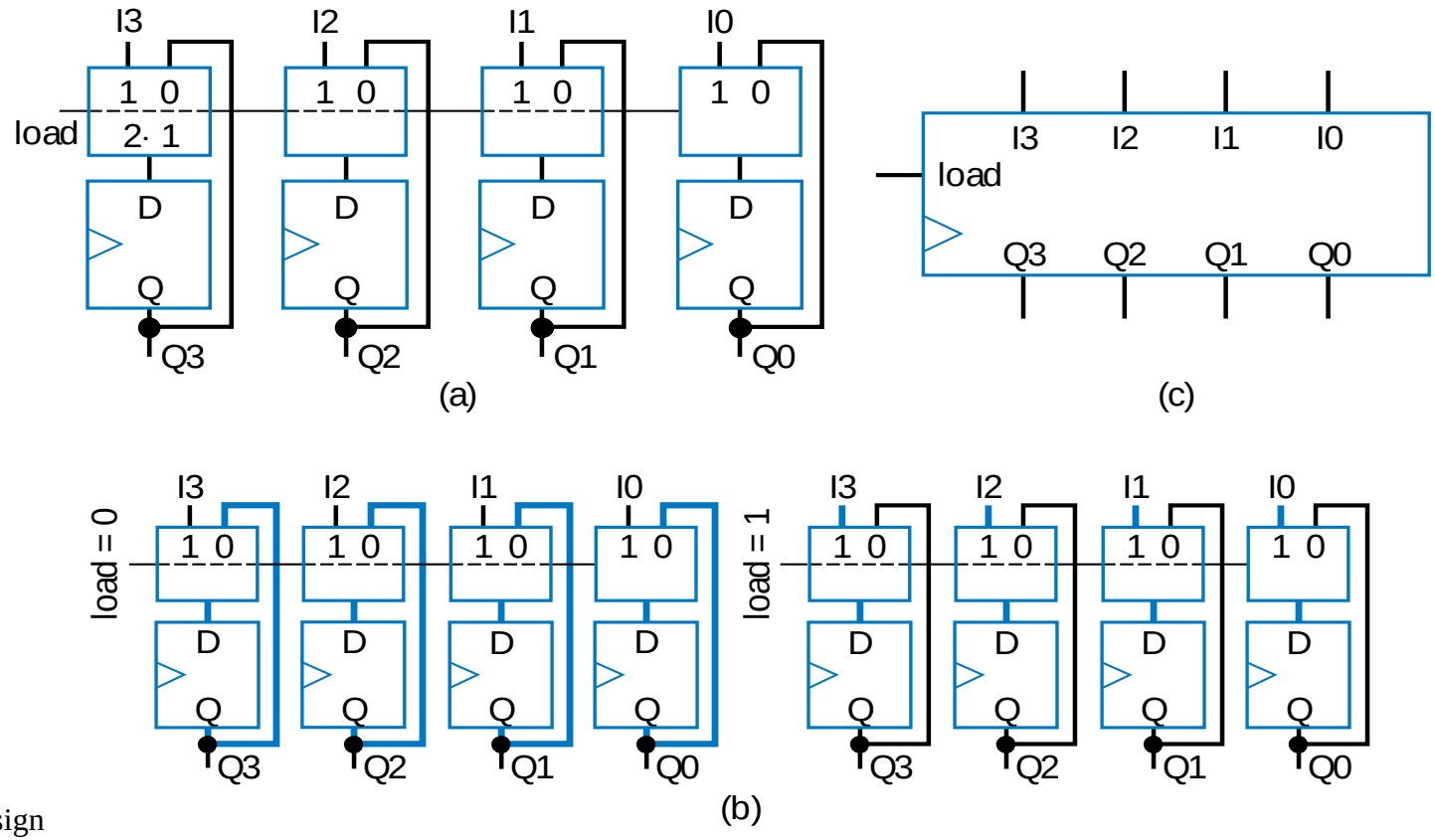


Basic register loads on every clock cycle

How extend to only load on certain cycles?

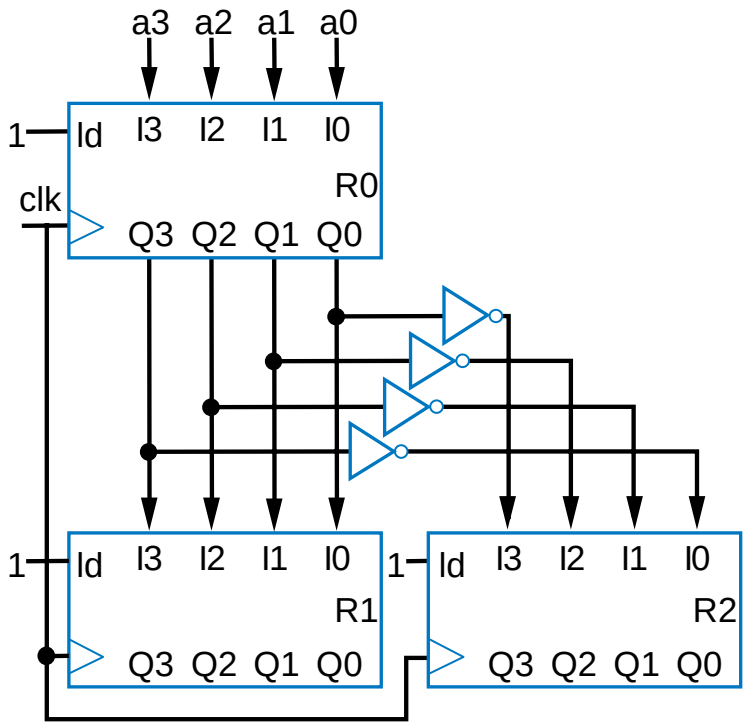
Register with Parallel Load

- Add 2x1 mux to front of each flip-flop
- Register's load input selects mux input to pass
  - Either existing flip-flop value, or new value to load

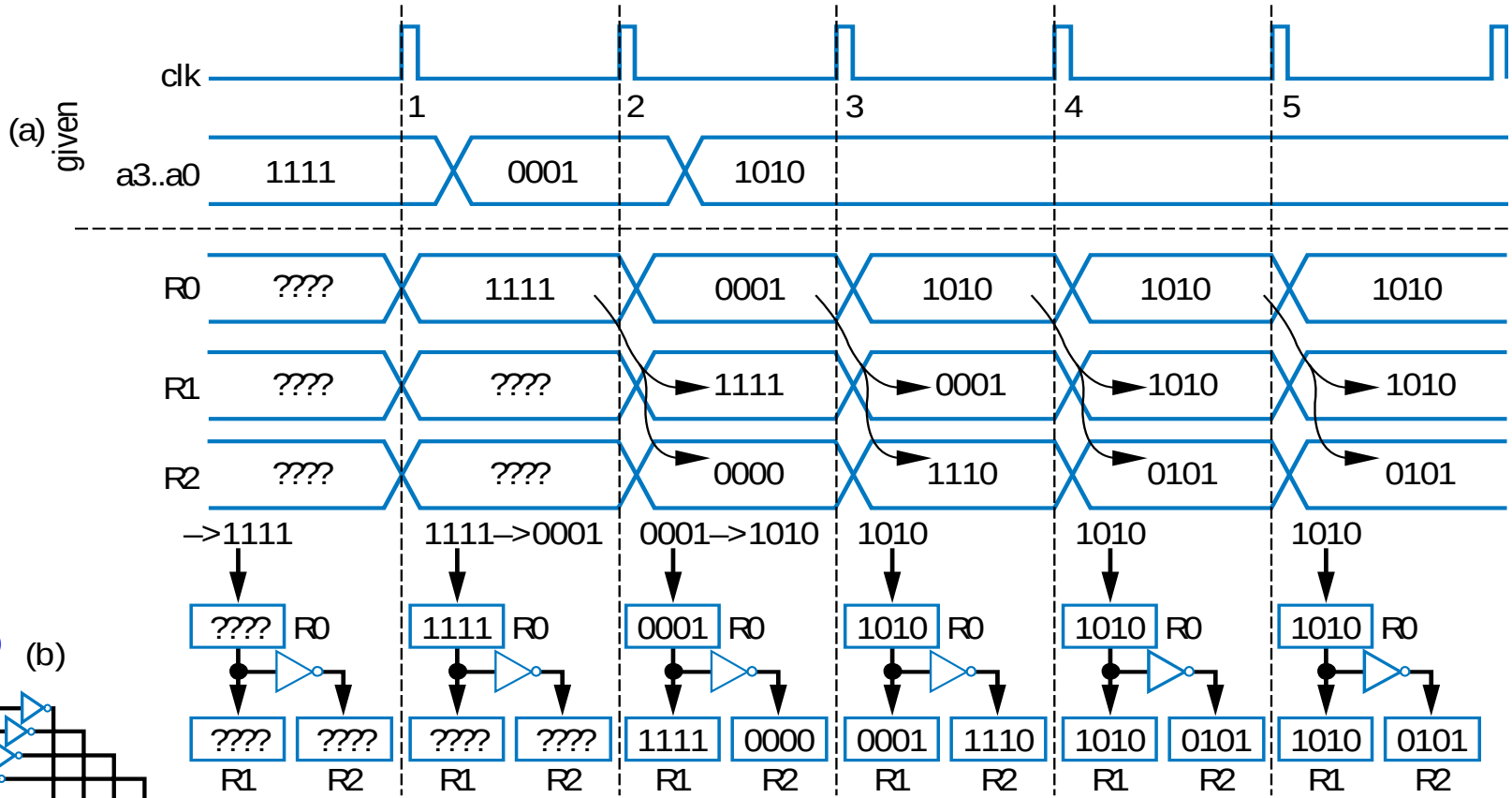


Basic Example Using Registers

- This example will show how registers load simultaneously on clock cycles
  - Notice that all load inputs set to 1 in this example -- just for demonstration purposes

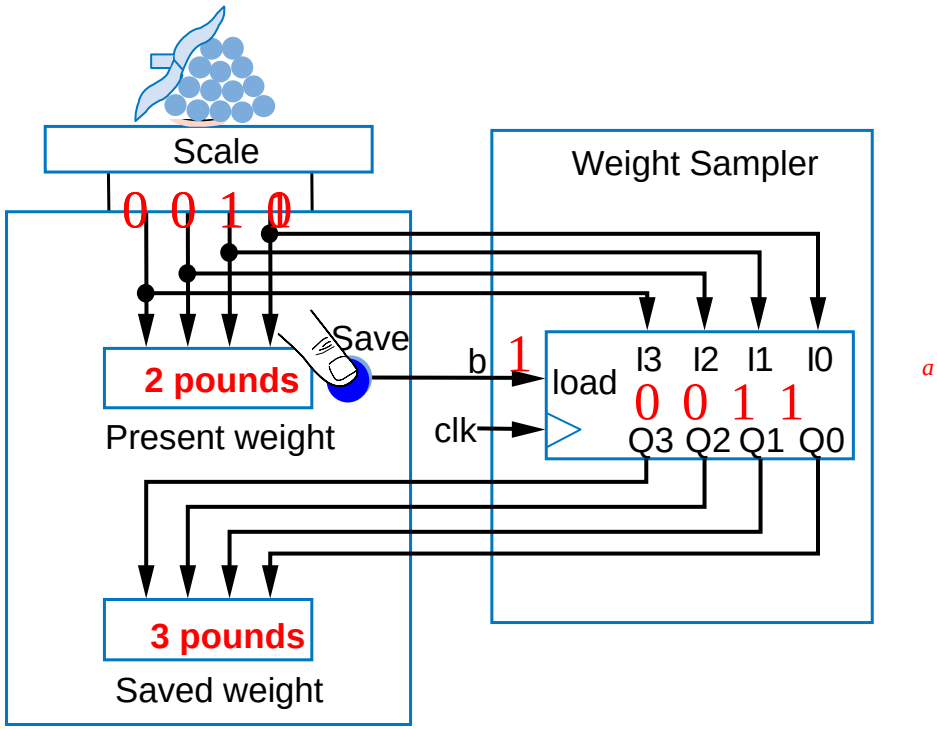


Basic Example Using Registers



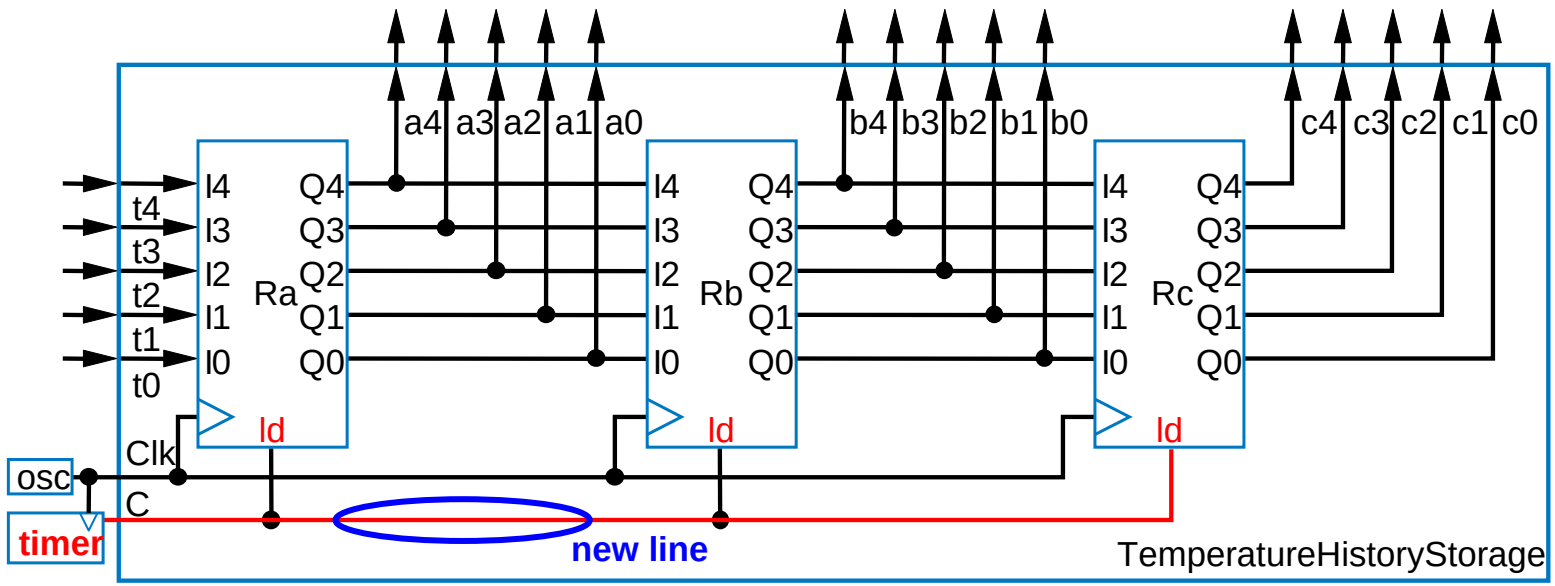
Register Example using the Load Input: Weight Sampler

- Scale has two displays
  - Present weight
  - Saved weight
  - Useful to compare present item with previous item
- Use register to store weight
  - Pressing button causes present weight to be stored in register
    - Register contents always displayed as “Saved weight,” even when new present weight appears

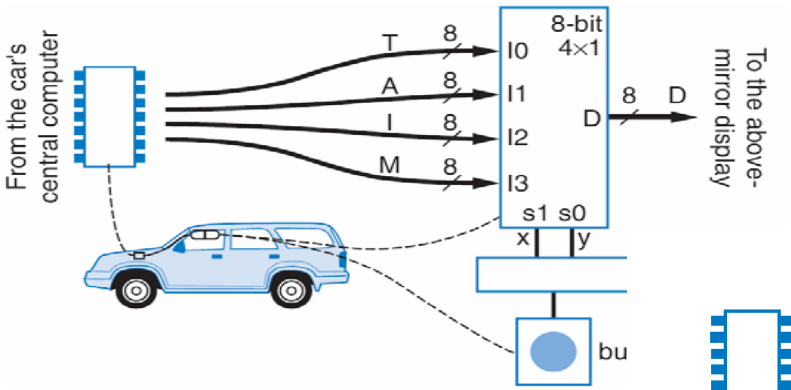


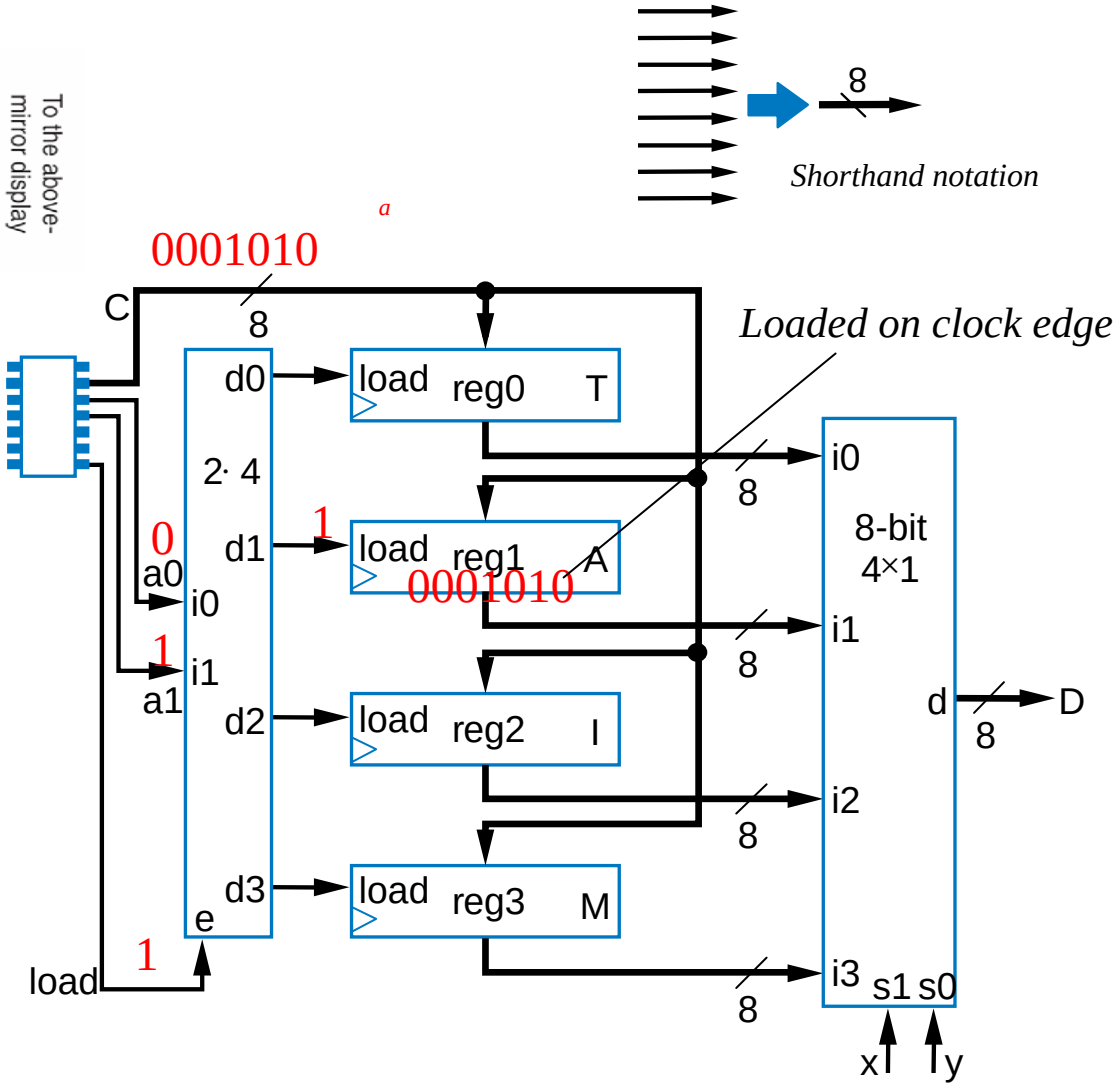
Register Example: Temperature History Display

- Recall Chpt 3 example
  - Timer pulse every hour
  - Previously used as clock. Better design only connects oscillator to clock inputs -- use registers with load input, connect to timer pulse.



Register Example: Above-Mirror Display

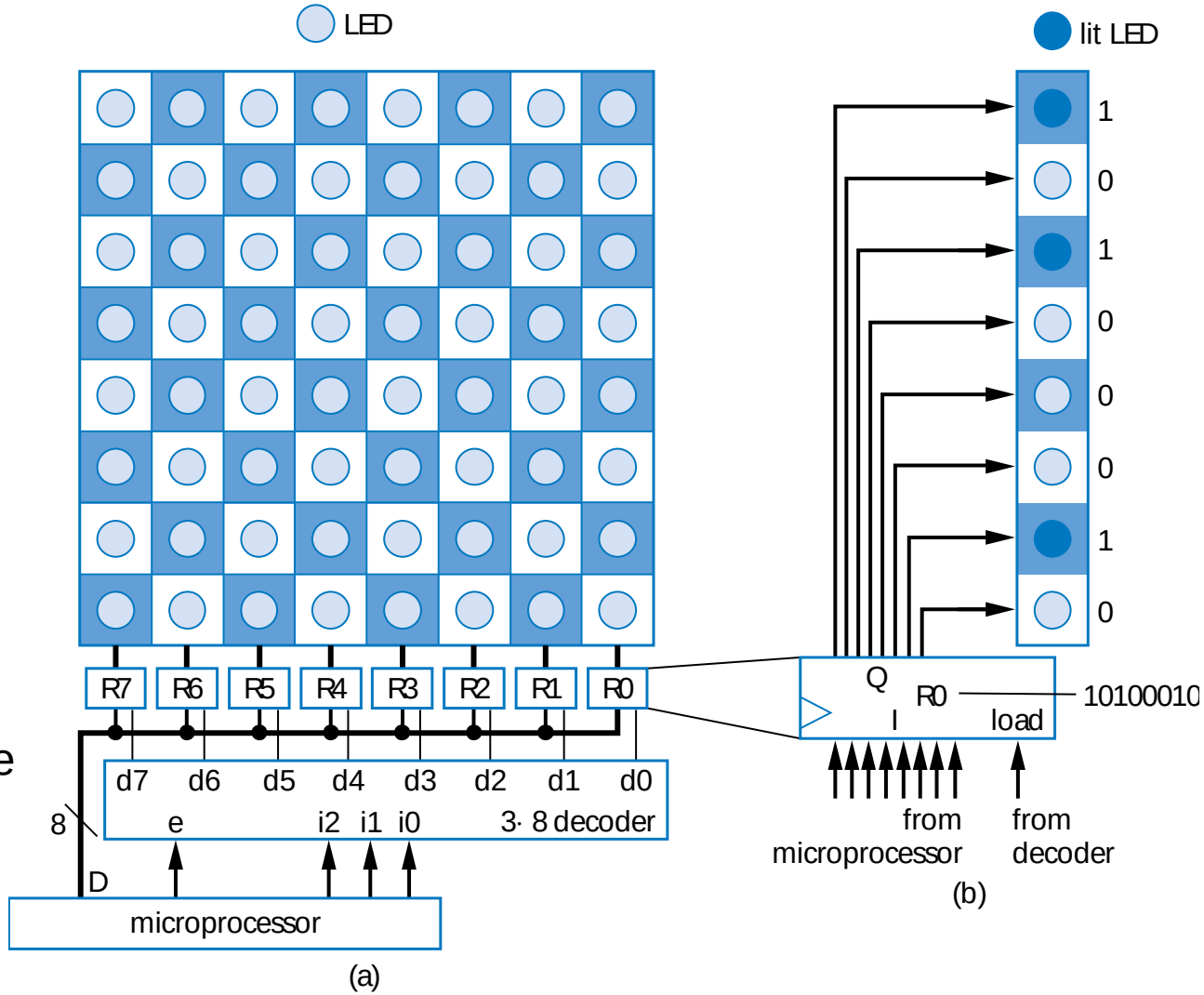




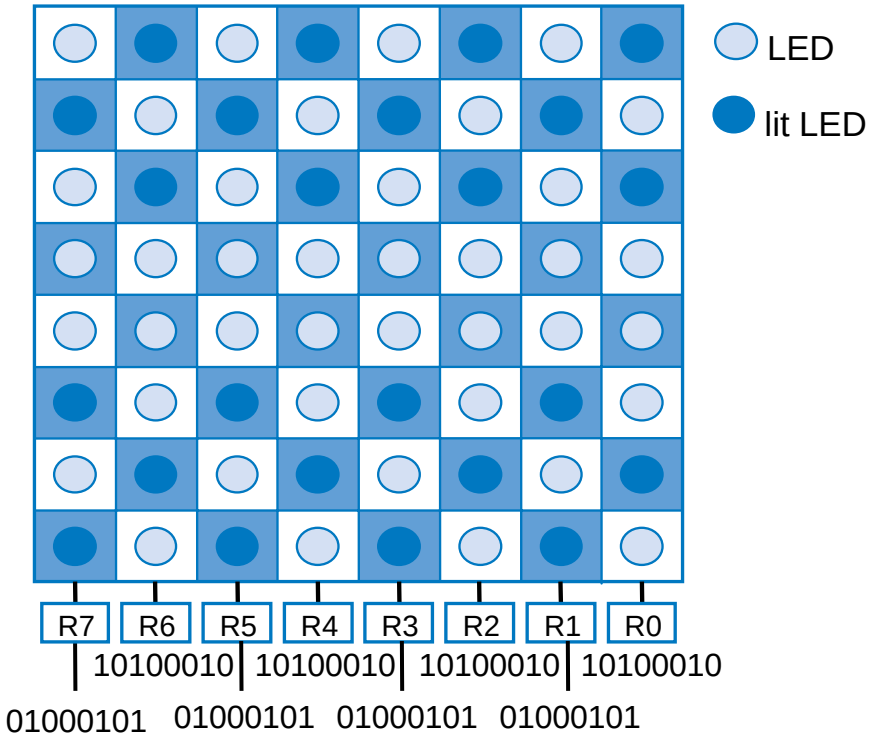
- Ch2 example: Four simultaneous values from car's computer
- To reduce wires: Computer writes only 1 value at a time, loads into one of four registers
  - Was: 8+8+8+8 = 32 wires
  - Now: 8 +2+1 = 11 wires

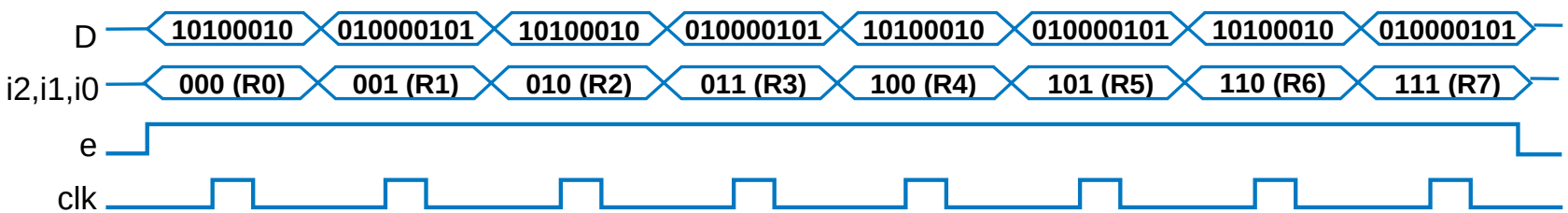
Register Example: Computerized Checkerboard

- Each register holds values for one column of lights
  - 1 lights light
- Microprocessor loads one register at a time
  - Occurs fast enough that user sees entire board change at once



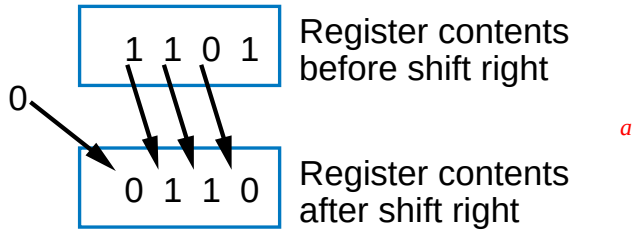
Register Example: Computerized Checkerboard





Shift Register

- Shift right
  - Move each bit one position right
  - Shift in 0 to leftmost bit



Q: Do four right shifts on 1001, showing value after each shift

A:

1001 (original)

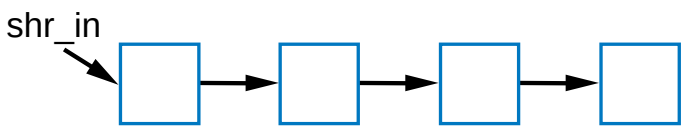
0100

0010

0001

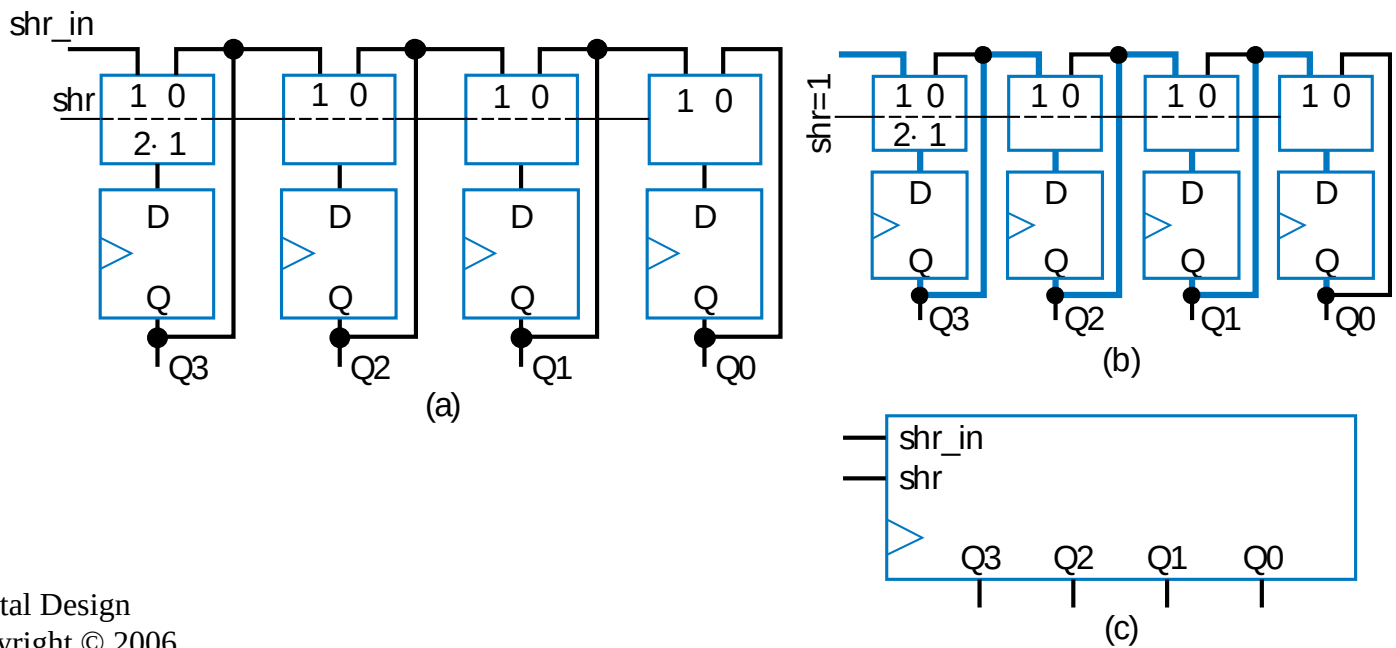
0000

- Implementation: Connect flip-flop output to next flip-flop's input



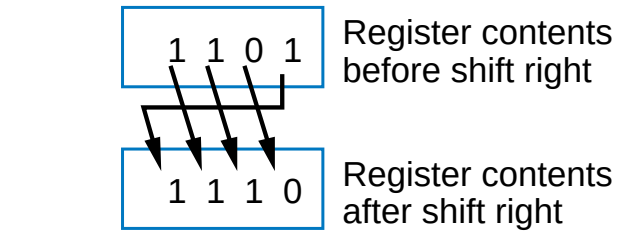
Shift Register

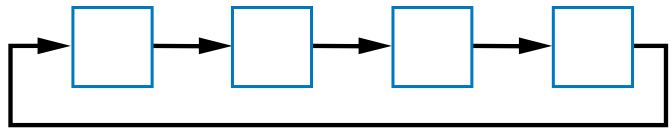
- To allow register to either shift or retain, use 2x1 muxes
  - shr: 0 means retain, 1 shift
  - shr_in: value to shift in
    - May be 0, or 1
- Note: Can easily design shift register that shifts left instead



Rotate Register

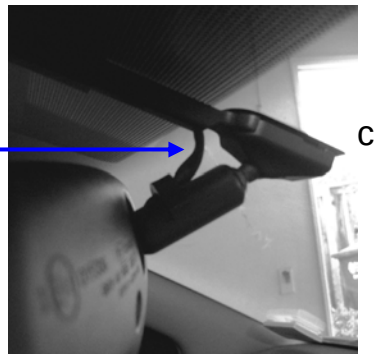
- Rotate right: Like shift right, but leftmost bit comes from rightmost bit

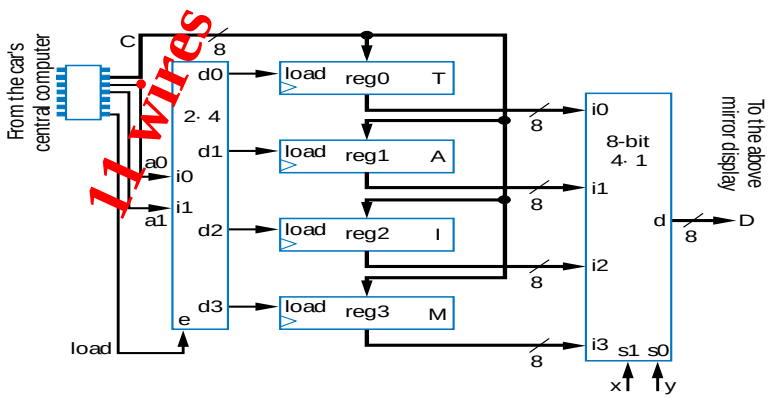


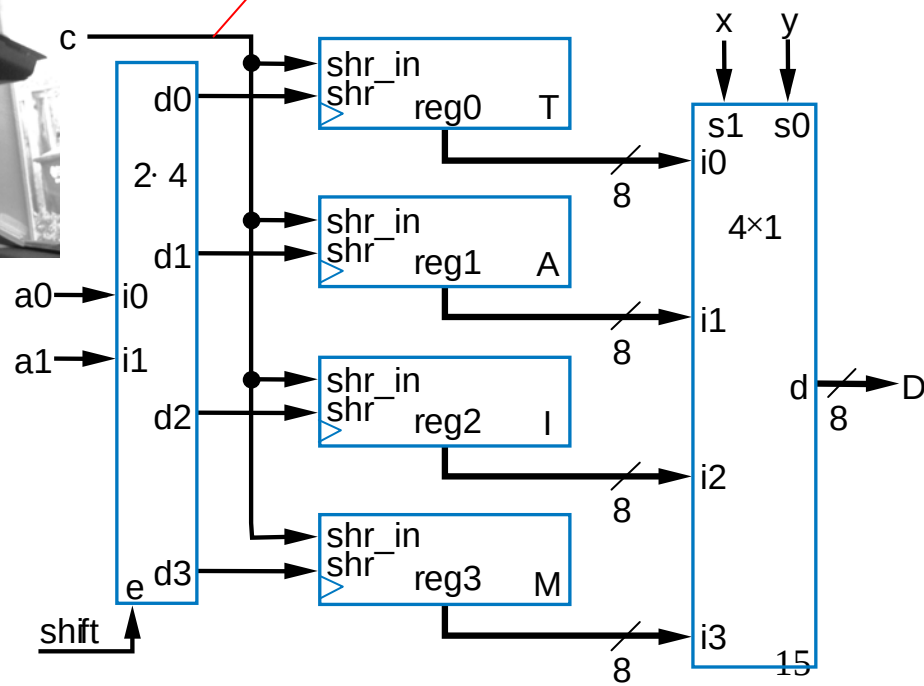


Shift Register Example: Above-Mirror Display

- Earlier example: 8 +2+1 = 11wires from car's computer to above-mirror display's four registers
  - Better than 32 wires, but 11 still a lot -- want fewer for smaller wire bundles
- Use shift registers
  - Wires: 1+2+1=4
  - Computer sends one value at a time, one bit per clock cycle





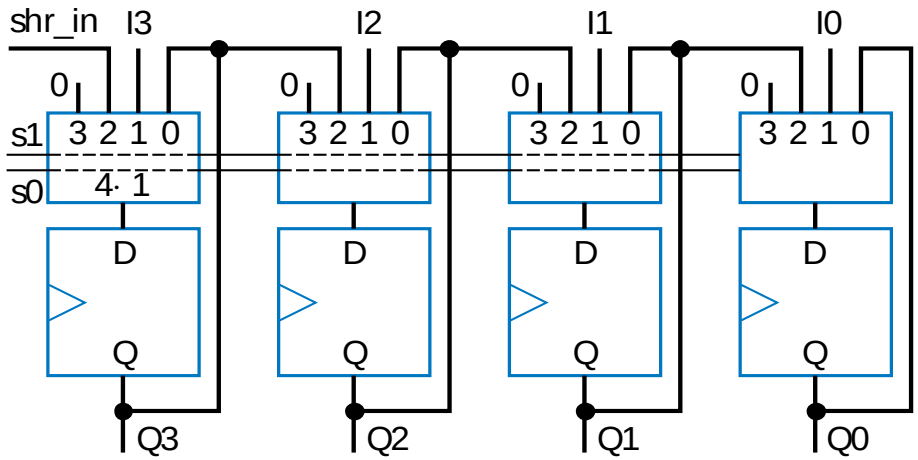


Multifunction Registers

- Many registers have multiple functions
  - Load, shift, clear (load all 0s)
  - And retain present value, of course
- Easily designed using muxes
  - Just connect each mux input to achieve desired function

Functions:

s1	s0	Operation
0	0	Maintain present value
0	1	Parallel load
1	0	Shift right
1	1	(unused - let's load 0s)



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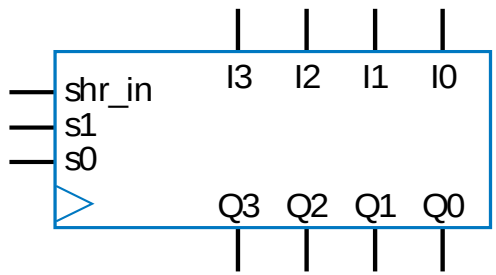
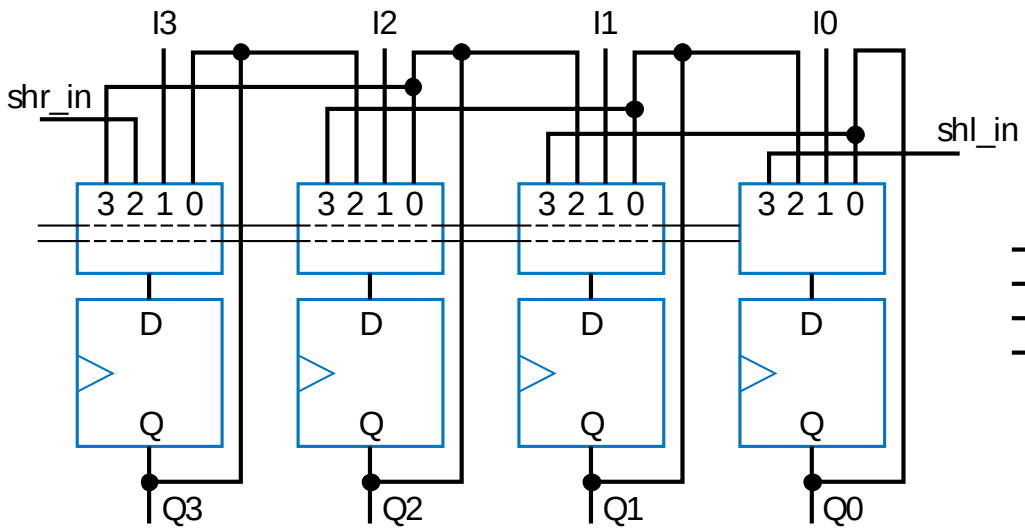


(b)

Multifunction Registers

s1	s0	Operation
0	0	Maintain present value
0	1	Parallel load
1	0	Shift right
1	1	Shift left



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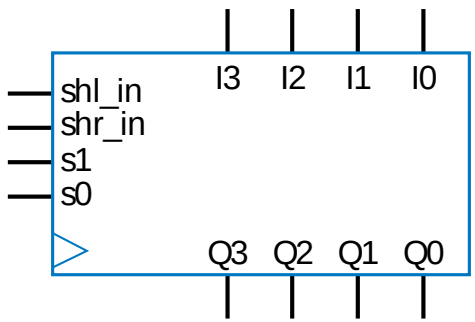
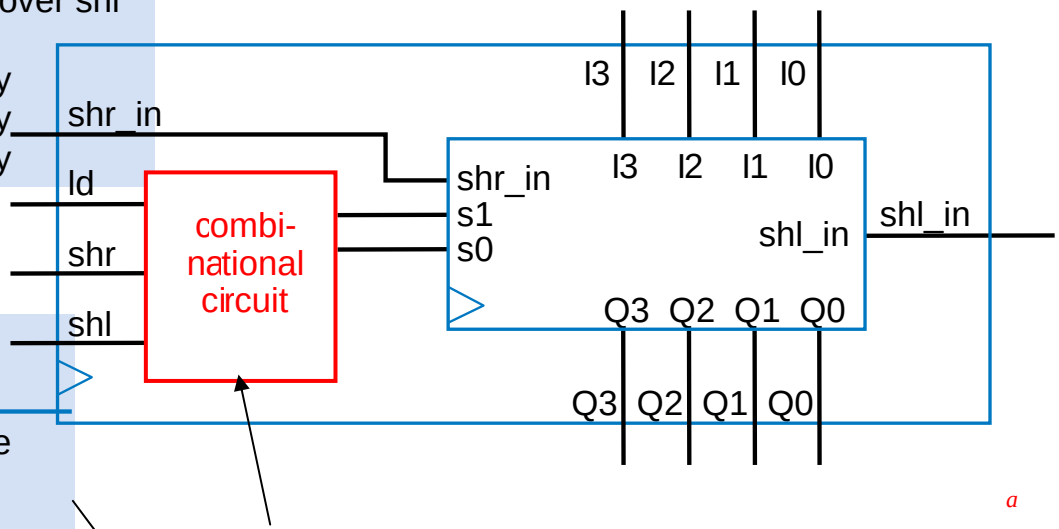


(b)

Multifunction Registers with Separate Control Inputs

ld	shr	shl	Operation
0	0	0	Maintain present value
0	0	1	Shift left
0	1	0	Shift right
0	1	1	Shift right – shr has priority over shl
1	0	0	Parallel load
1	0	1	Parallel load – ld has priority
1	1	0	Parallel load – ld has priority
1	1	1	Parallel load – ld has priority



Truth table for combinational circuit

Inputs ld	shr	shl	Outputs s1	s0	Note Operation
0	0	0	0	0	Maintain value
0	0	1	1	1	Shift left
0	1	0	1	0	Shift right
0	1	1	1	0	Shift right
1	0	0	0	1	Parallel load
1	0	1	0	1	Parallel load
1	1	0	0	1	Parallel load
1	1	1	0	1	Parallel load

$s1 = ld'*shr'*shl + ld'*shr*shl' + ld'*shr*shl$

$s0 = ld'*shr'*shl + ld$

Register Operation Table

- Register operations typically shown using compact version of table
  - X means same operation whether value is 0 or 1
    - One X expands to two rows
    - Two Xs expand to four rows
  - Put highest priority control input on left to make reduced table simple

Inputs			Outputs		Note
ld	shr	shl	s1	s0	Operation
0	0	0	0	0	Maintain value
0	0	1	1	1	Shift left
0	1	**0**	1	0	Shift right
0	1	**1**	1	0	Shift right
1	**0**	**0**	0	1	Parallel load
1	**0**	**1**	0	1	Parallel load
1	**1**	**0**	0	1	Parallel load
1	**1**	**1**	0	1	Parallel load

ld	shr	shl	Operation
0	0	0	Maintain value
0	0	1	Shift left
0	1	**X**	Shift right
1	**X**	**X**	Parallel load

Register Design Process

- Can design register with desired operations using simple four-step process

TABLE 4.1 Four-step process for designing a multifunction register.

	Step	Description
1.	*Determine mux size*	Count the number of operations (don't forget the maintain present value operation!) and add in front of each flip-flop a mux with at least that number of inputs.
2.	*Create mux operation table*	Create an operation table defining the desired operation for each possible value of the mux select lines.
3.	*Connect mux inputs*	For each operation, connect the corresponding mux data input to the appropriate external input or flip-flop output (possibly passing through some logic) to achieve the desired operation.
4.	*Map control lines*	Create a truth table that maps external control lines to the internal mux select lines, with appropriate priorities, and then design the logic to achieve that mapping

Register Design Example

- Desired register operations
  - Load, shift left, synchronous clear, synchronous set

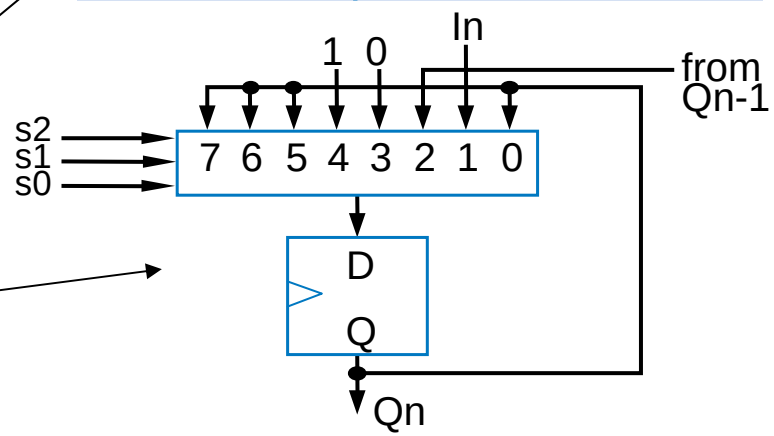
s2	s1	s0	Operation
0	0	0	Maintain present value
0	0	1	Parallel load
0	1	0	Shift left
0	1	1	Synchronous clear
1	0	0	Synchronous set
1	0	1	Maintain present value
1	1	0	Maintain present value
1	1	1	Maintain present value

Step 1: Determine mux size

5 operations: above, plus maintain present value (don't forget this one!)
--> **Use 8x1 mux**

Step 2: Create mux operation table

Step 3: Connect mux inputs

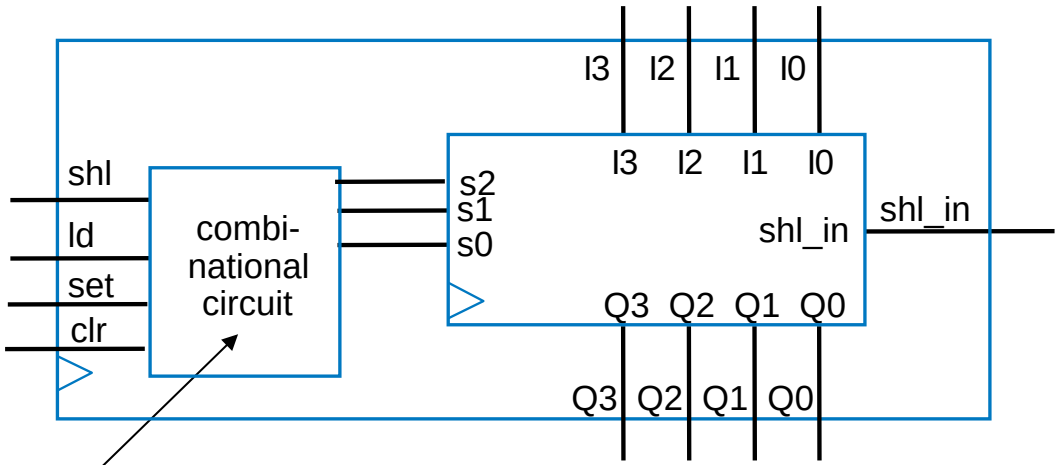


Step 4: Map control lines

s2 = clr'*set
s1 = clr'*set'*ld'*shl + clr
s0 = clr'*set'*ld + clr

Inputs				Outputs			
clr	set	ld	shl	s2	s1	s0	Operation
0	0	0	0	0	0	0	Maintain present value
0	0	0	1	0	1	0	Shift left
0	0	1	X	0	0	1	Parallel load
0	1	X	X	1	0	0	Set to all 1s
1	X	X	X	0	1	1	Clear to all 0s

Register Design Example



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s2 = clr'*set
s1 = clr'*set'*ld'*shl + clr
s0 = clr'*set'*ld + clr

Inputs				Outputs			
clr	set	ld	shl	s2	s1	s0	Operation
0	0	0	0	0	0	0	Maintain present value
0	0	0	1	0	1	0	Shift left
0	0	1	X	0	0	1	Parallel load
0	1	X	X	1	0	0	Set to all 1s
1	X	X	X	0	1	1	Clear to all 0s

Adders

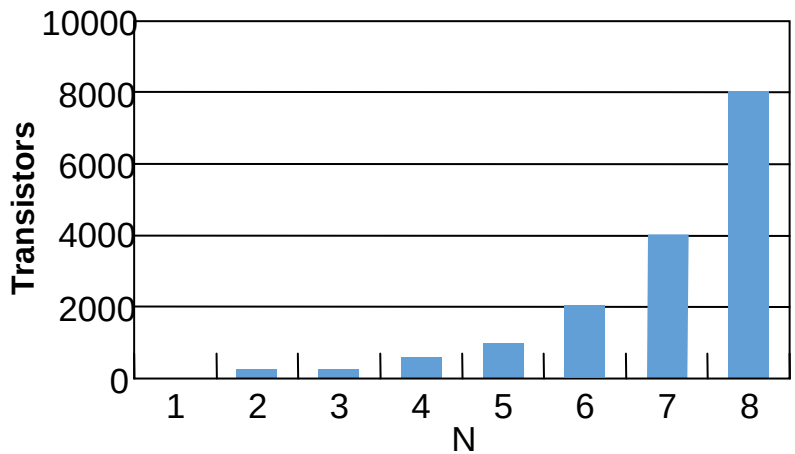
- Adds two N-bit binary numbers
  - 2-bit adder: adds two 2-bit numbers, outputs 3-bit result
  - e.g., 01 + 11 = 100 (1 + 3 = 4)
- Can design using combinational design process of Ch 2, but doesn't work well for reasonable-size N
  - Why not?

Inputs				Outputs		
a1	a0	b1	b0	c	s1	s0
0	0	0	0	0	0	0
0	0	0	1	0	0	1
0	0	1	0	0	1	0
0	0	1	1	0	1	1
0	1	0	0	0	0	1
0	1	0	1	0	1	0
0	1	1	0	0	1	1
0	1	1	1	1	0	0
1	0	0	0	0	1	0
1	0	0	1	0	1	1
1	0	1	0	1	0	0
1	0	1	1	1	0	1
1	1	0	0	0	1	1
1	1	0	1	1	0	0
1	1	1	0	1	0	1
1	1	1	1	1	1	0

Why Adders Aren't Built Using Standard Combinational Design Process

- Truth table too big
  - 2-bit adder's truth table shown
    - Has $2^{(2+2)}$ = 16 rows
  - 8-bit adder: $2^{(8+8)}$ = 65,536 rows
  - 16-bit adder: $2^{(16+16)}$ = ~4 billion rows
  - 32-bit adder: ...
- Big truth table with numerous 1s/0s yields big logic
  - Plot shows number of transistors for N-bit adders, using state-of-the-art automated combinational design tool

Inputs				Outputs		
a1	a0	b1	b0	c	s1	s0
0	0	0	0	0	0	0
0	0	0	1	0	0	1
0	0	1	0	0	1	0
0	0	1	1	0	1	1
0	1	0	0	0	0	1
0	1	0	1	0	1	0
0	1	1	0	0	1	1
0	1	1	1	1	0	0
1	0	0	0	0	1	0
1	0	0	1	0	1	1
1	0	1	0	1	0	0
1	0	1	1	1	0	1
1	1	0	0	0	1	1
1	1	0	1	1	0	0
1	1	1	0	1	0	1
1	1	1	1	1	1	0

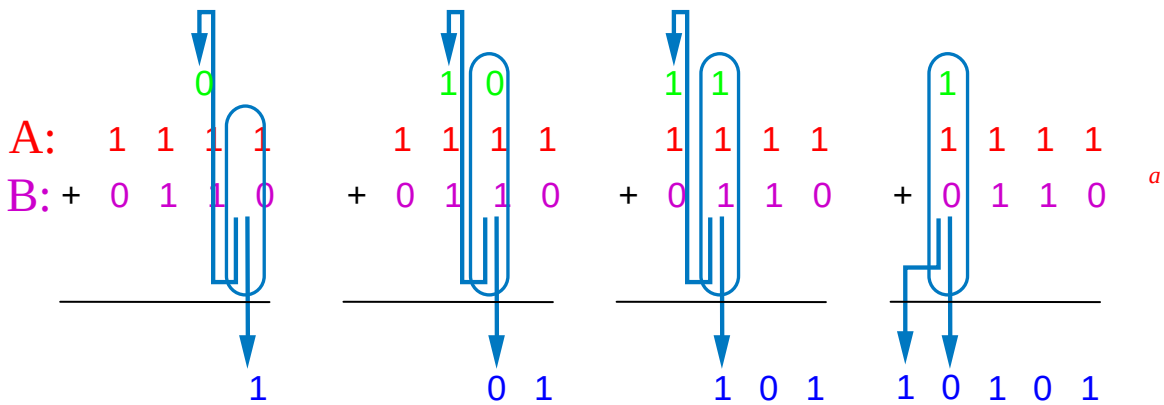


Q: Predict number of transistors for 16-bit adder

A: 1000 transistors for N=5, doubles for each increase of N. So transistors = $1000*2^{(N-5)}$. Thus, for N=16, transistors = $1000*2^{(16-5)} = 1000*2048 = 2{,}048{,}000$. Way too many!

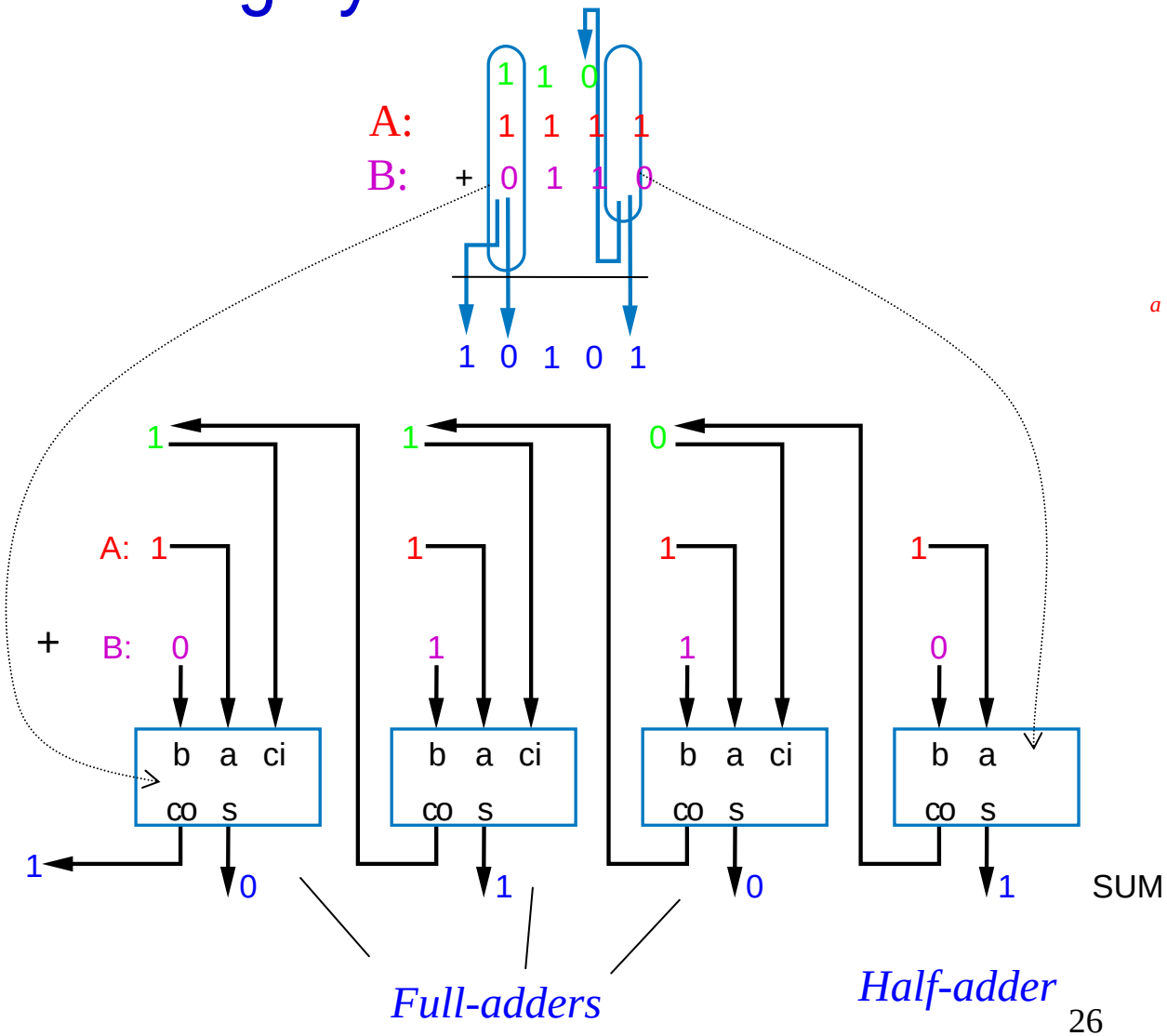
Alternative Method to Design an Adder: Imitate Adding by Hand

- Alternative adder design: mimic how people do addition by hand
- One column at a time
  - Compute sum, add carry to next column



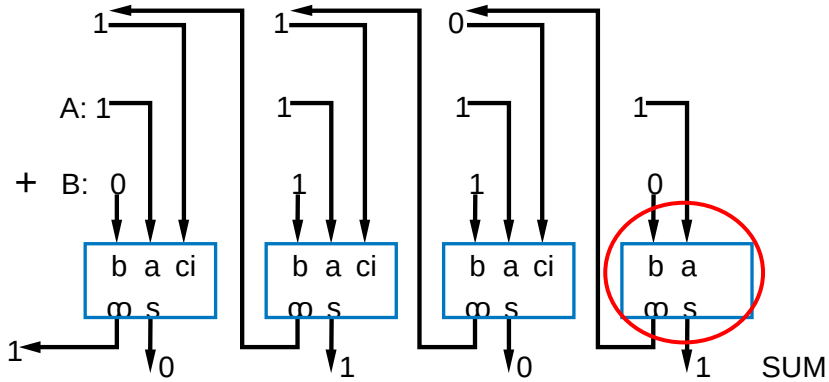
Alternative Method to Design an Adder: Imitate Adding by Hand

- Create component for each column
  - Adds that column's bits, generates sum and carry bits



Half-Adder

- ***Half-adder***: Adds 2 bits, generates sum and carry
- Design using combinational design process from Ch 2



Step 1: Capture the function

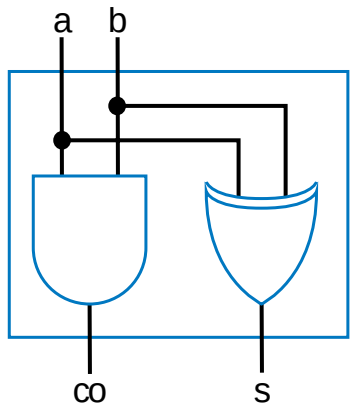
Inputs		Outputs	
a	b	co	s
0	0	0	0
0	1	0	1
1	0	0	1
1	1	1	0

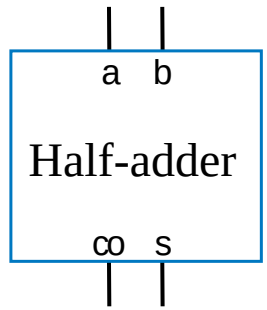
Step 2: Convert to equations

co = ab

s = a'b + ab' (same as s = a xor b)

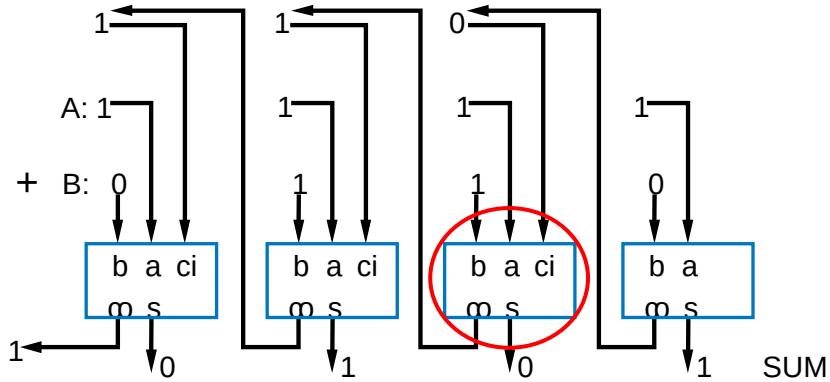
Step 3: Create the circuit





Full-Adder

- ***Full-adder:*** Adds 3 bits, generates sum and carry
- Design using combinational design process from Ch 2



Step 1: Capture the function

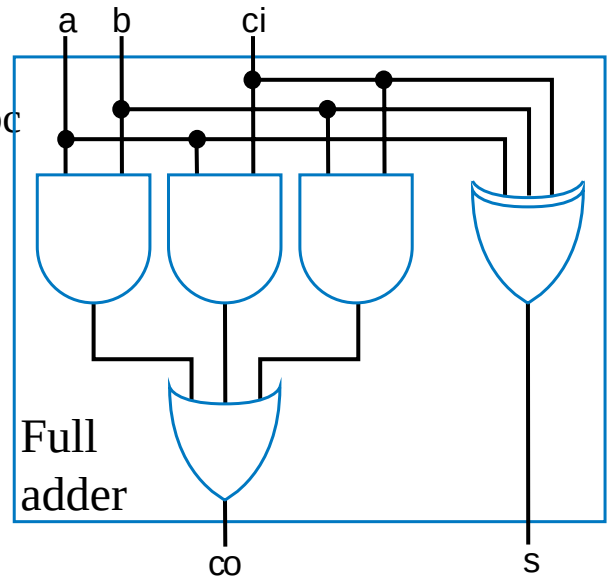
Inputs			Outputs	
a	b	ci	co	s
0	0	0	0	0
0	0	1	0	1
0	1	0	0	1
0	1	1	1	0
1	0	0	0	1
1	0	1	1	0
1	1	0	1	0
1	1	1	1	1

Step 2: Convert to equations

co = a'bc + ab'c + abc' + abc
co = a'bc +abc +ab'c +abc +abc' +abc
co = (a'+a)bc + (b'+b)ac + (c'+c)ab
co = bc + ac + ab

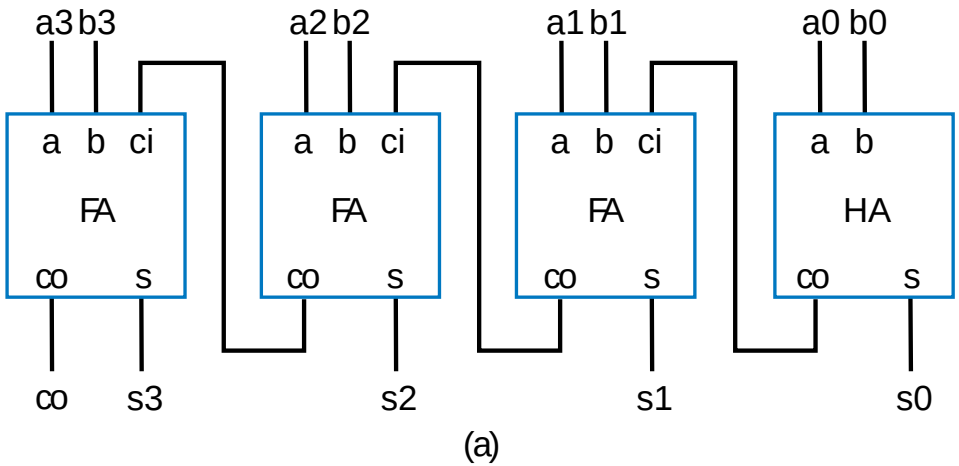
s = a'b'c + a'bc' + ab'c' + abc
s = a'(b'c + bc') + a(b'c' + bc)
s = a'(b xor c)' + a(b xor c)
s = a xor b xor c

Step 3: Create the circuit

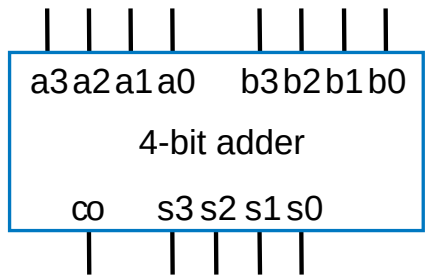


Carry-Ripple Adder

- Using half-adder and full-adders, we can build adder that adds like we would by hand
- Called a ***carry-ripple adder***
  - 4-bit adder shown: Adds two 4-bit numbers, generates 5-bit output
    - 5-bit output can be considered 4-bit "sum" plus 1-bit "carry out"
  - Can easily build any size adder



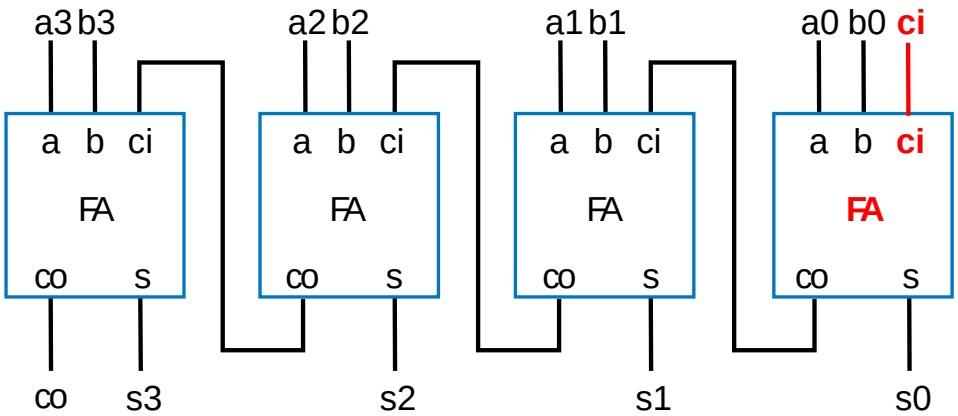
(a)



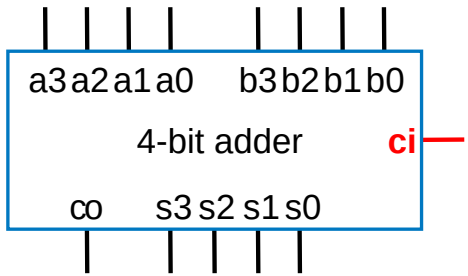
(b)

Carry-Ripple Adder

- Using full-adder instead of half-adder for first bit, we can include a “carry in” bit in the addition
  - Will be useful later when we connect smaller adders to form bigger adders

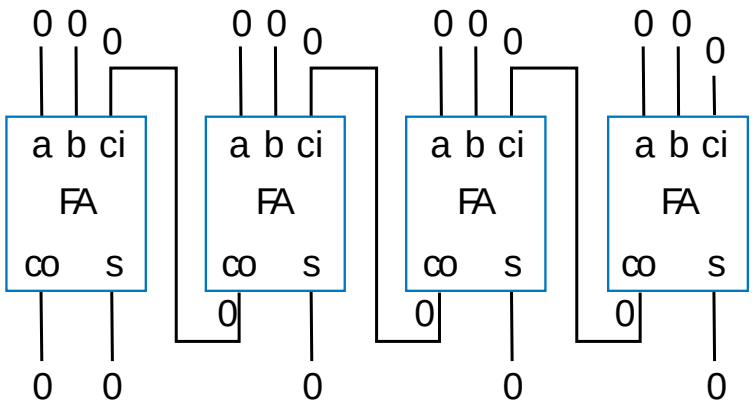


(a)

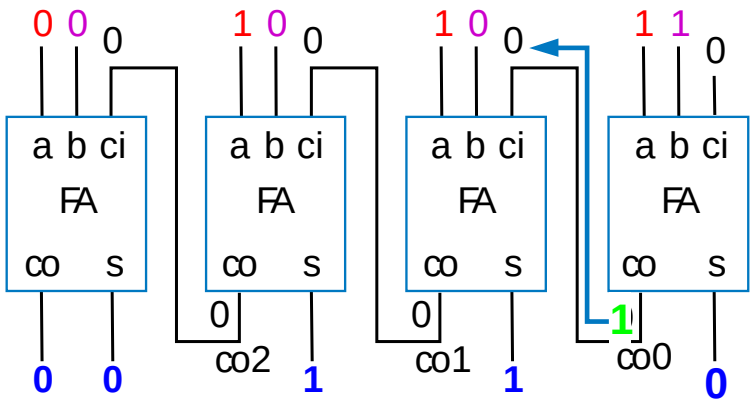


(b)

Carry-Ripple Adder's Behavior



Assume all inputs initially 0

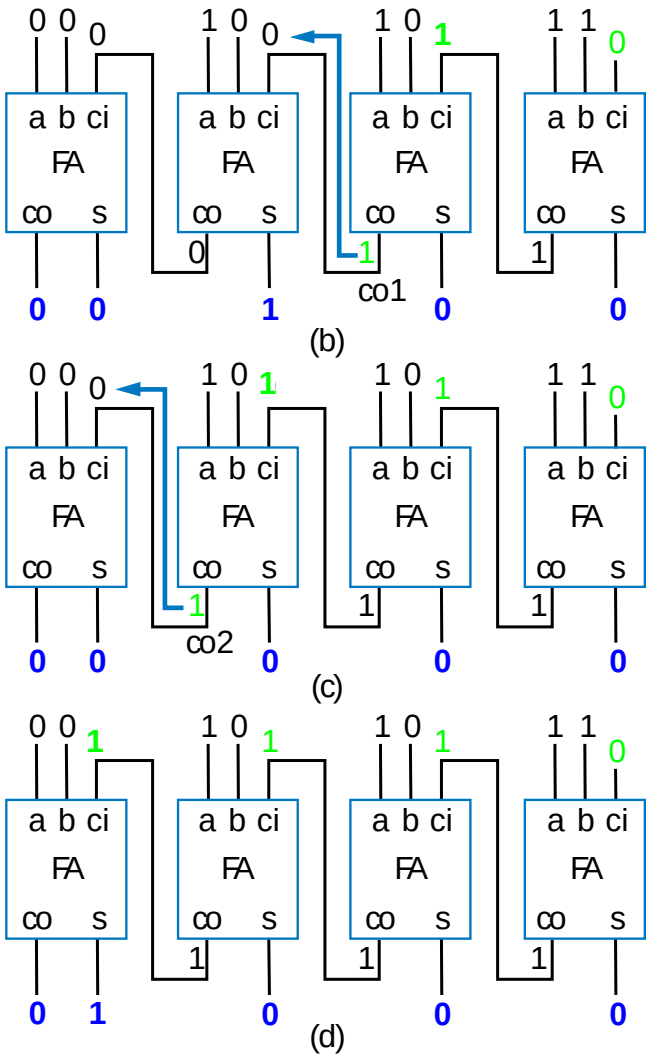


0111+0001
(answer should be 01000)

Output after 2 ns (1FA delay)

Wrong answer -- something wrong? No -- just need more time for carry to ripple through the chain of full adders.

Carry-Ripple Adder's Behavior



0111+0001

(answer should be 01000)

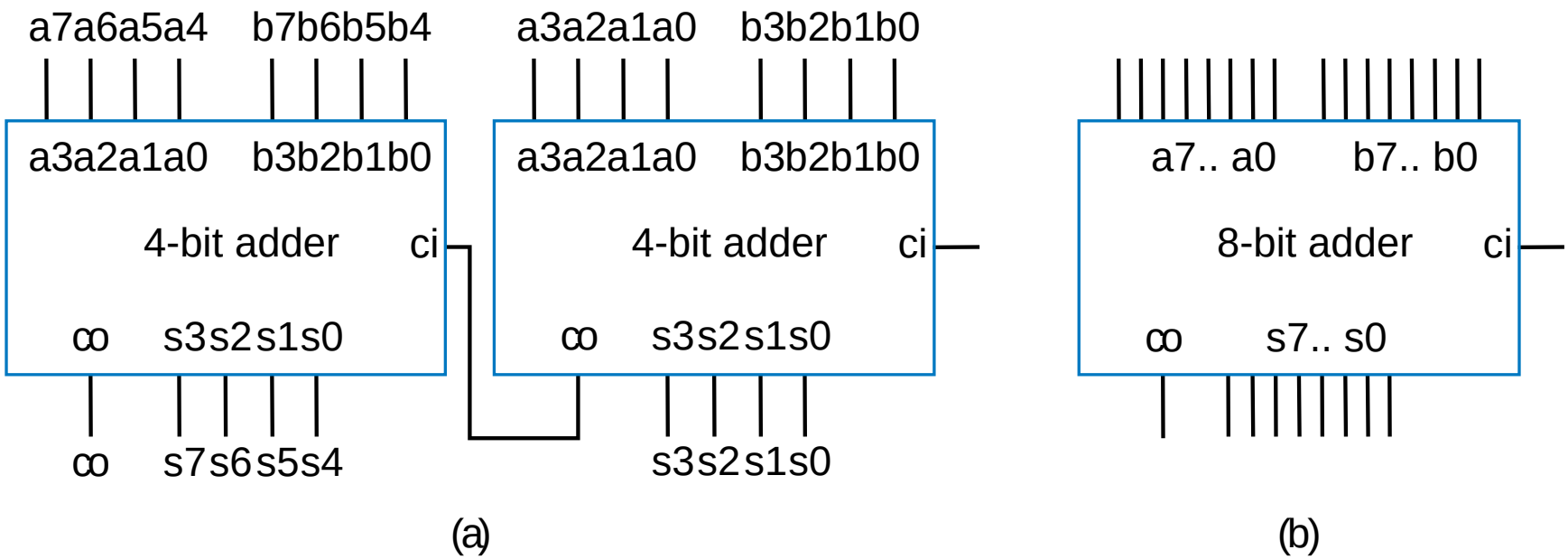
Outputs after 4ns (2 FA delays)

Outputs after 6ns (3 FA delays)

Output after 8ns (4 FA delays)

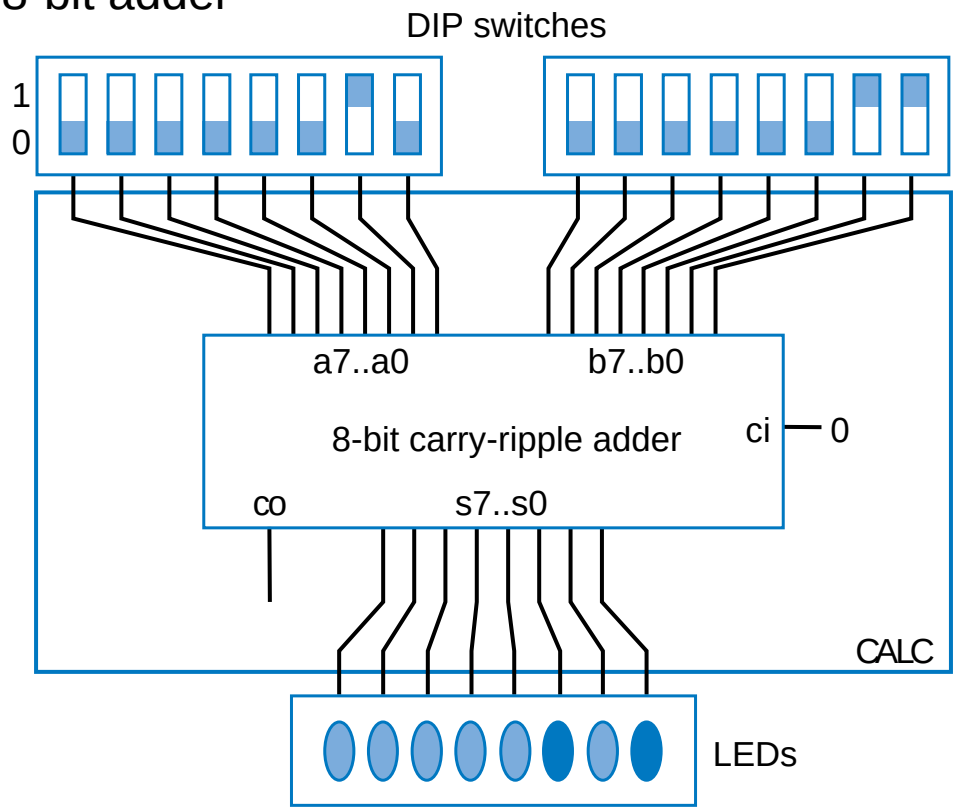
Correct answer appears after 4 FA delays

Cascading Adders



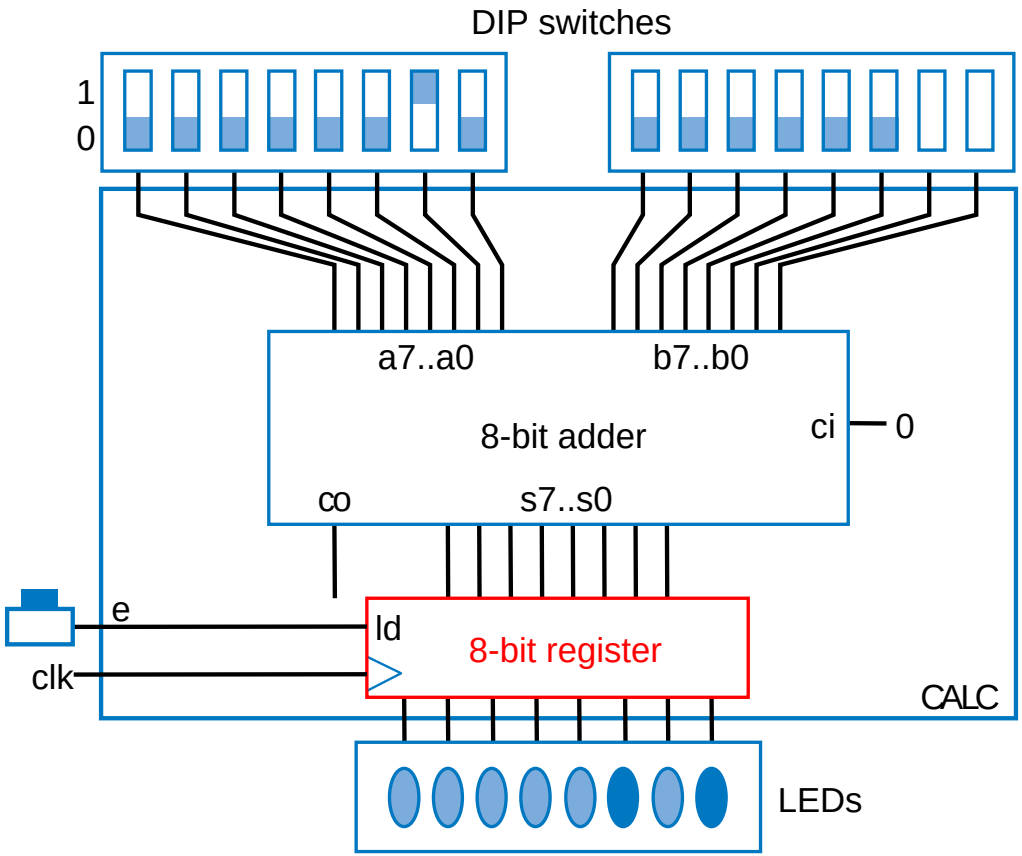
Adder Example: DIP-Switch-Based Adding Calculator

- Goal: Create calculator that adds two 8-bit binary numbers, specified using DIP switches
  - DIP switch: Dual-inline package switch, move each switch up or down
  - Solution: Use 8-bit adder



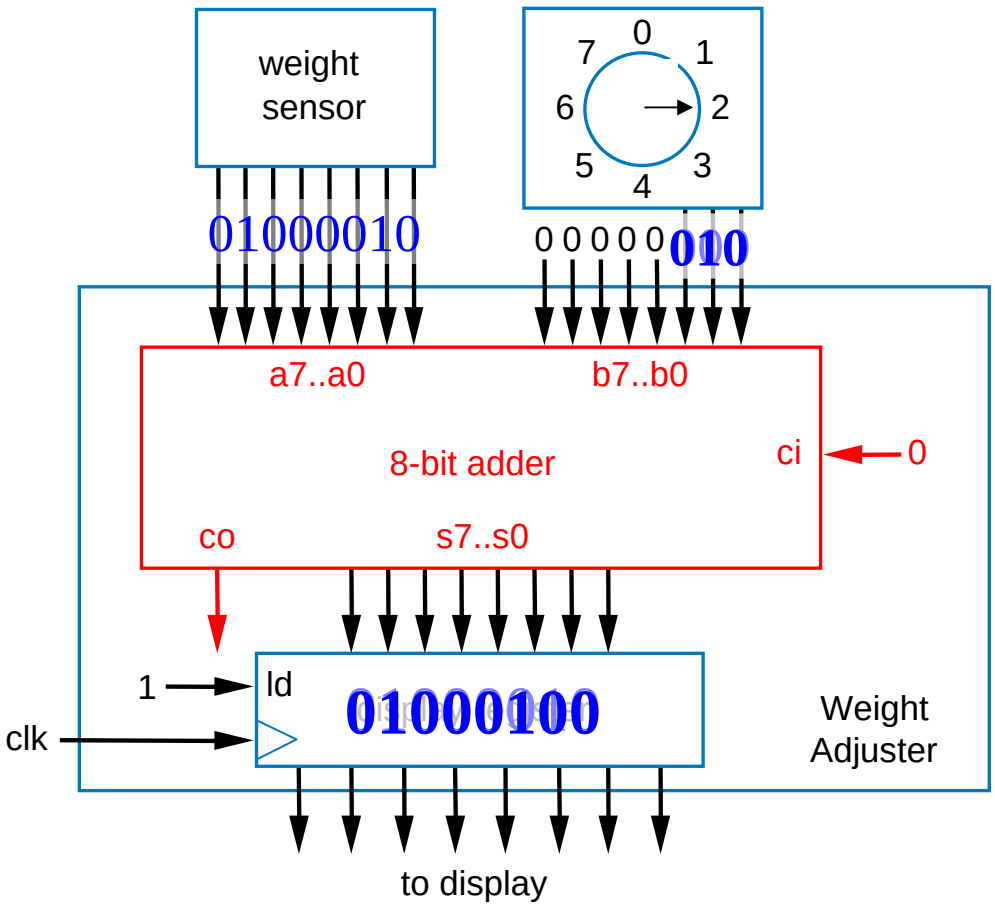
Adder Example: DIP-Switch-Based Adding Calculator

- To prevent spurious values from appearing at output, can place register at output
  - Actually, the light flickers from spurious values would be too fast for humans to detect -- but the principle of registering outputs to avoid spurious values being read by external devices (which normally aren't humans) applies here.



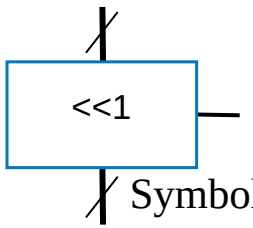
Adder Example: Compensating Weight Scale

- Weight scale with compensation amount of 0-7
  - To compensate for inaccurate sensor due to physical wear
  - Use 8-bit adder

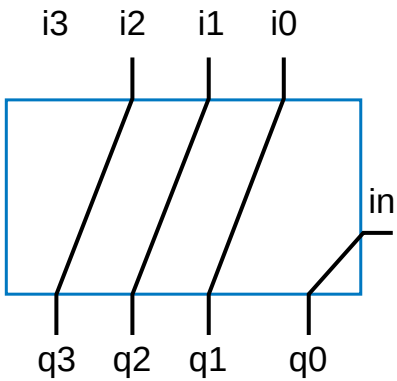


Shifters

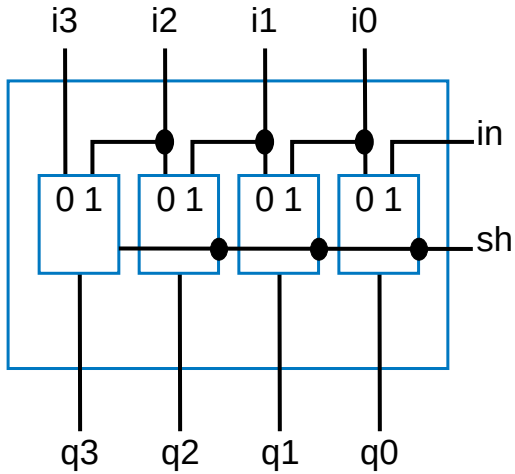
- Shifting (e.g., left shifting 0011 yields 0110) useful for:
  - Manipulating bits
  - Converting serial data to parallel (remember earlier above-mirror display example with shift registers)
  - Shift left once is same as multiplying by 2 (0011 (3) becomes 0110 (6))
    - Why? Essentially appending a 0 -- Note that multiplying decimal number by 10 accomplished just be appending 0, i.e., by shifting left (55 becomes 550)
  - Shift right once same as dividing by 2



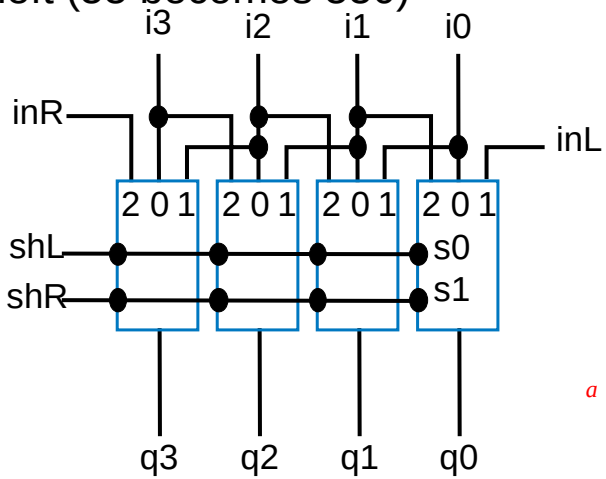
Symbol



Left shifter



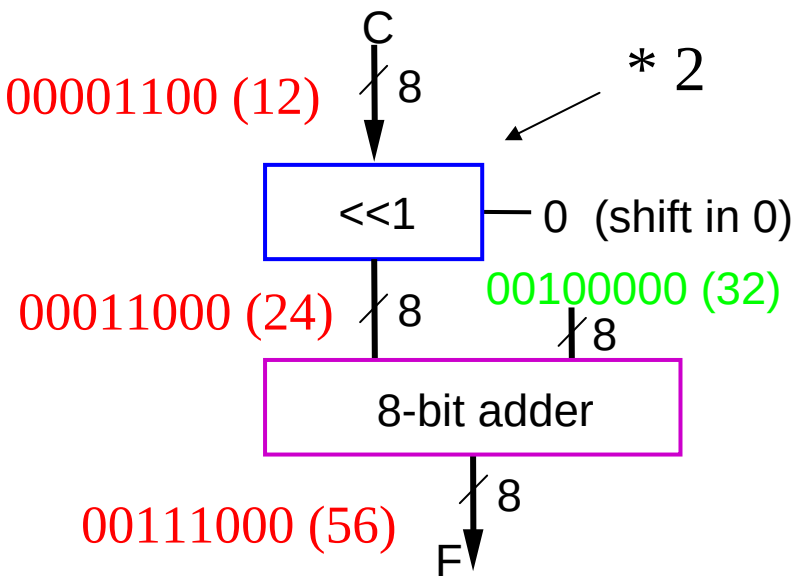
Shifter with left shift or no shift



Shifter with left shift, right shift, and no shift

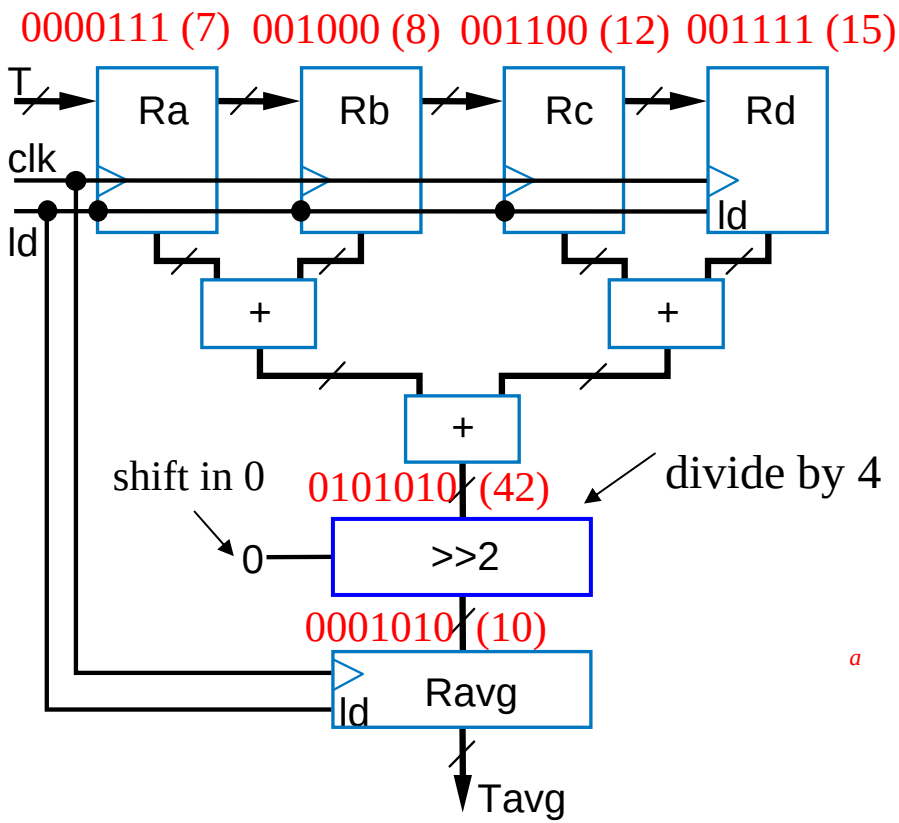
Shifter Example: Approximate Celsius to Fahrenheit Converter

- Convert 8-bit Celsius input to 8-bit Fahrenheit output
  - F = C * 9/5 + 32
  - Approximate: F = C*2 + 32
  - Use left shift: F = left_shift(C) + 32



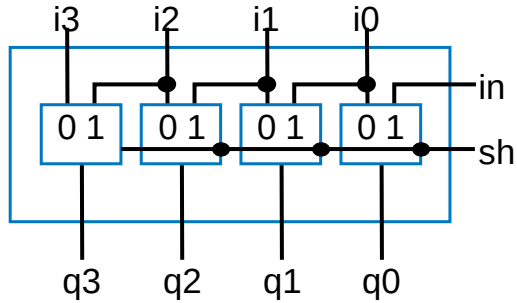
Shifter Example: Temperature Averager

- Four registers storing a history of temperatures
- Want to output the average of those temperatures
- Add, then divide by four
  - Same as shift right by 2
  - Use three adders, and right shift by two

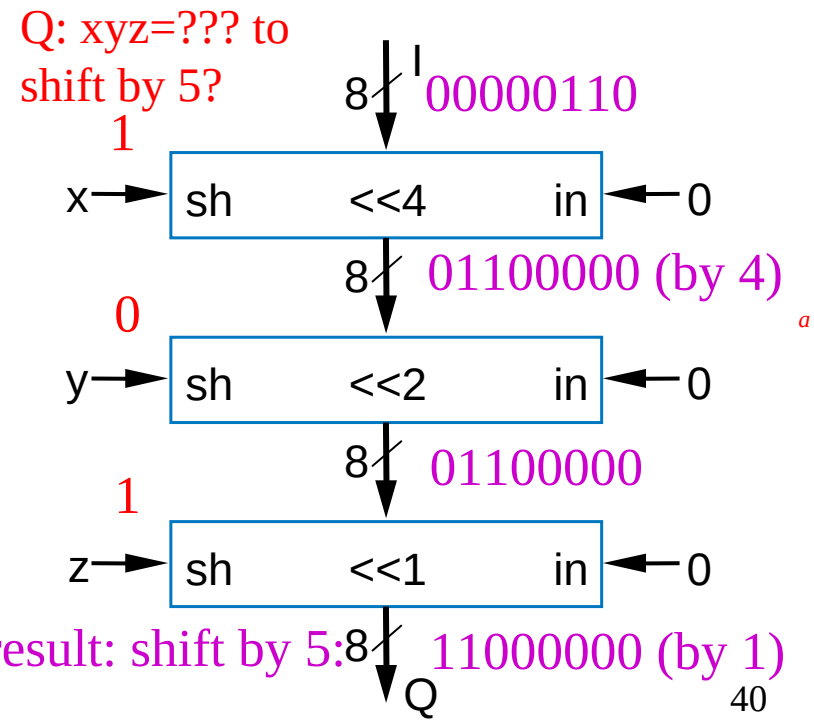


Barrel Shifter

- A shifter that can shift by any amount
  - 4-bit barrel left shift can shift left by 0, 1, 2, or 3 positions
  - 8-bit barrel left shifter can shift left by 0, 1, 2, 3, 4, 5, 6, or 7 positions
    - (Shifting an 8-bit number by 8 positions is pointless -- you just lose all the bits)
- Could design using 8x1 muxes and lots of wires
  - Too many wires
- More elegant design
  - Chain three shifters: 4, 2, and 1
  - Can achieve any shift of 0..7 by enabling the correct combination of those three shifters, i.e., shifts should sum to desired amount



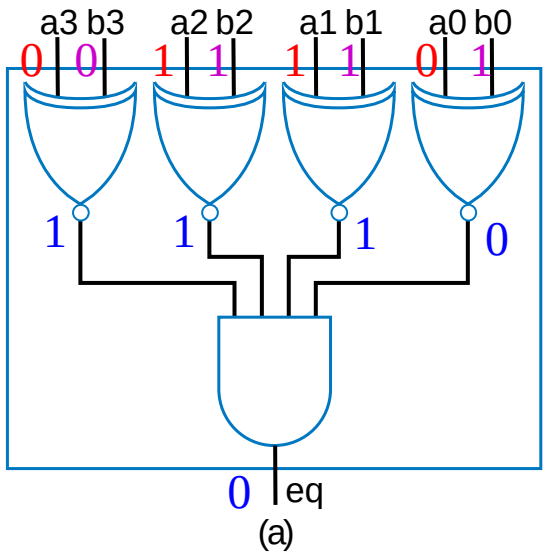
Shift by 1 shifter uses 2x1 muxes. 8x1 mux solution for 8-bit barrel shifter: too many wires.

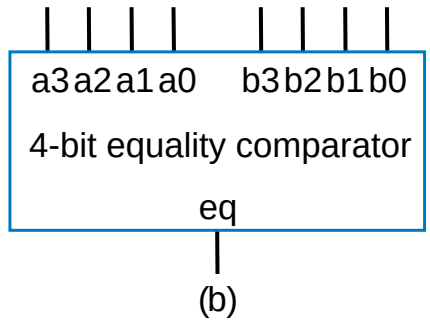


Comparators

- ***N-bit equality comparator***: Outputs 1 if two N-bit numbers are equal
  - 4-bit equality comparator with inputs A and B
    - a3 must equal b3, a2 = b2, a1 = b1, a0 = b0
      - Two bits are equal if both 1, or both 0
      - eq = (a3b3 + a3'b3') * (a2b2 + a2'b2') * (a1b1 + a1'b1') * (a0b0 + a0'b0')
    - Recall that XNOR outputs 1 if its two input bits are the same
      - eq = (a3 xnor b3) * (a2 xnor b2) * (a1 xnor b1) * (a0 xnor b0)

0110 = 0111 ?





Magnitude Comparator

- ***N-bit magnitude comparator***: Indicates whether A>B, A=B, or A<B, for its two N-bit inputs A and B
  - How design? Consider how compare by hand. First compare a3 and b3. If equal, compare a2 and b2. And so on. Stop if comparison not equal -- whichever's bit is 1 is greater. If never see unequal bit pair, A=B.

A=1011 B=1001

1011 **1**001 Equal

1**0**11 1**0**01 Equal

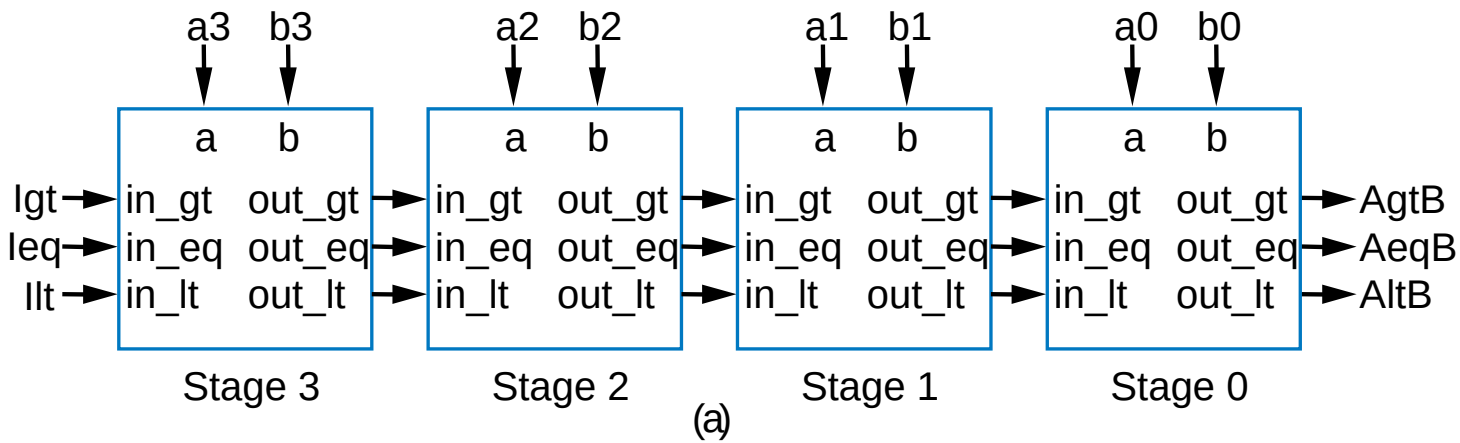
10**1**1 10**0**1 **Unequal**

So A > B

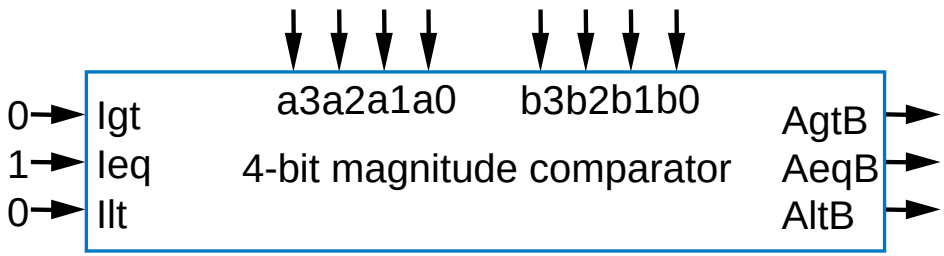
a

Magnitude Comparator

- By-hand example leads to idea for design
  - Start at left, compare each bit pair, pass results to the right
  - Each bit pair called a stage
  - Each stage has 3 inputs indicating results of higher stage, passes results to lower stage

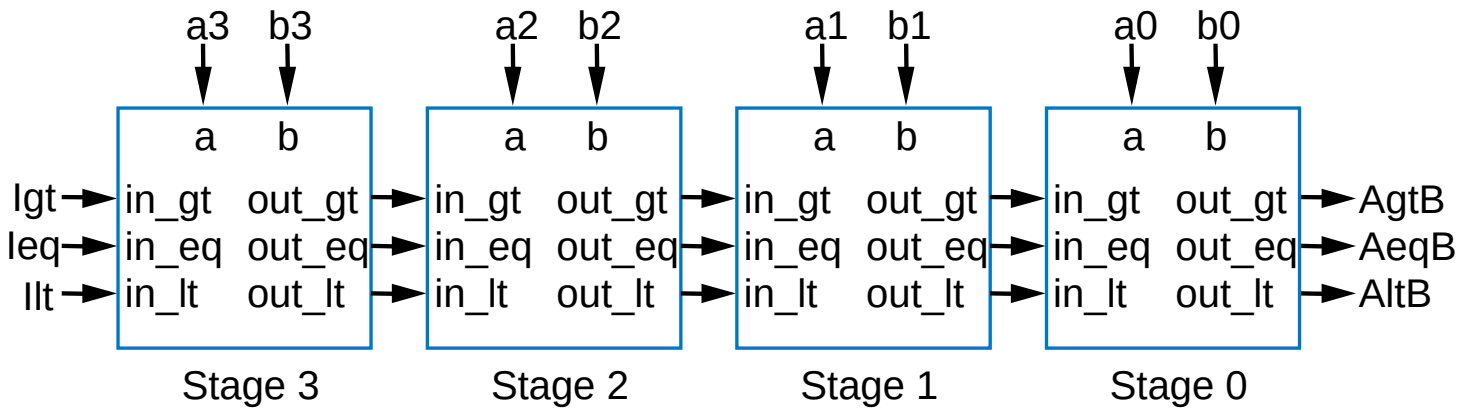


(a)



(b)

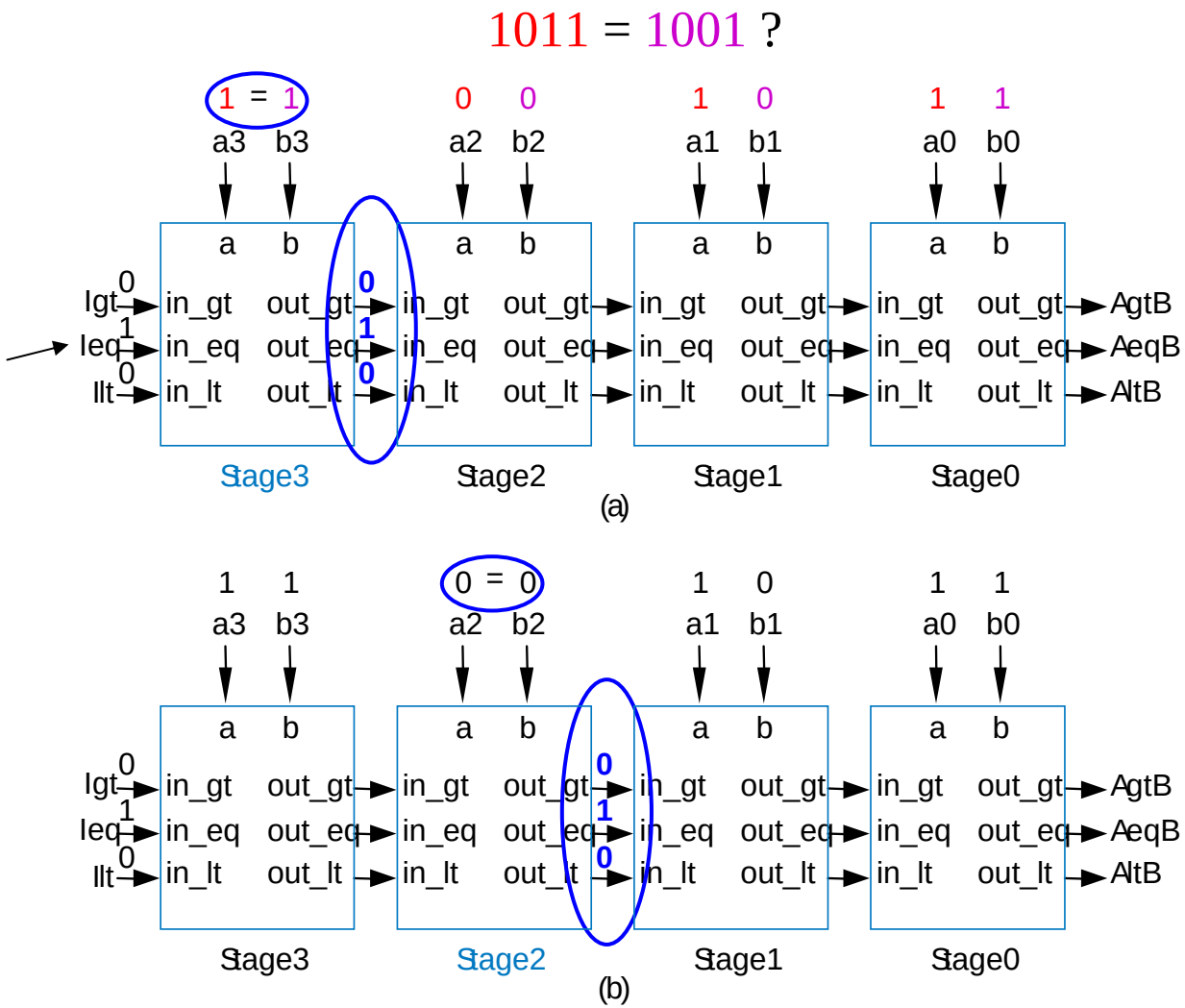
Magnitude Comparator



- Each stage:
  - out_gt = in_gt + (in_eq * a * b')
    - A>B (so far) if already determined in higher stage, or if higher stages equal but in this stage a=1 and b=0
  - out_lt = in_lt + (in_eq * a' * b)
    - A<B (so far) if already determined in higher stage, or if higher stages equal but in this stage a=0 and b=1
  - out_eq = in_eq * (a XNOR b)
    - A=B (so far) if already determined in higher stage and in this stage a=b too
  - Simple circuit inside each stage, just a few gates (not shown)

Magnitude Comparator

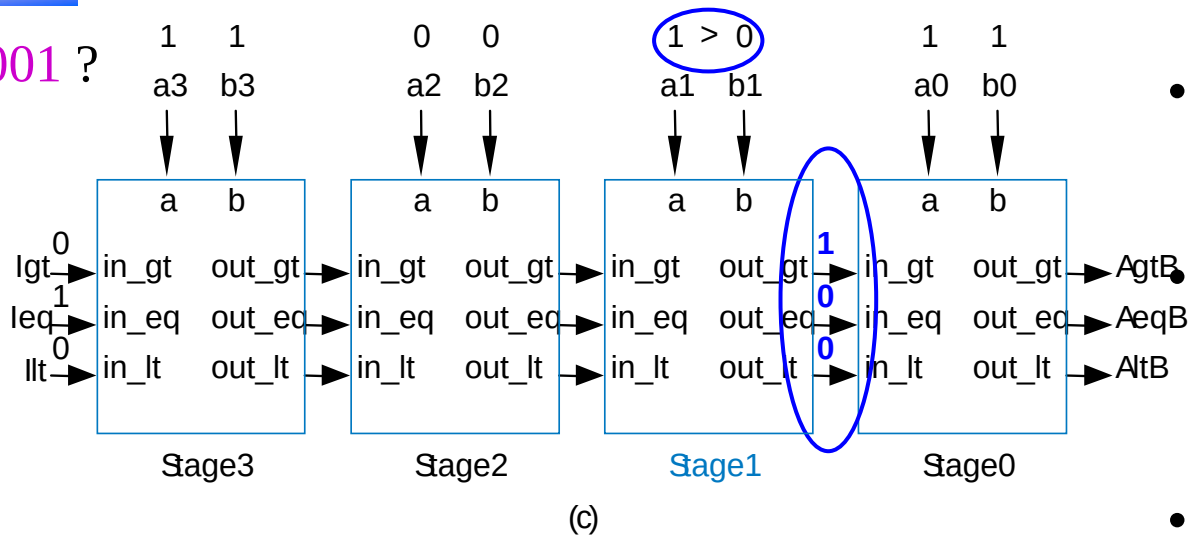
- How does it work?



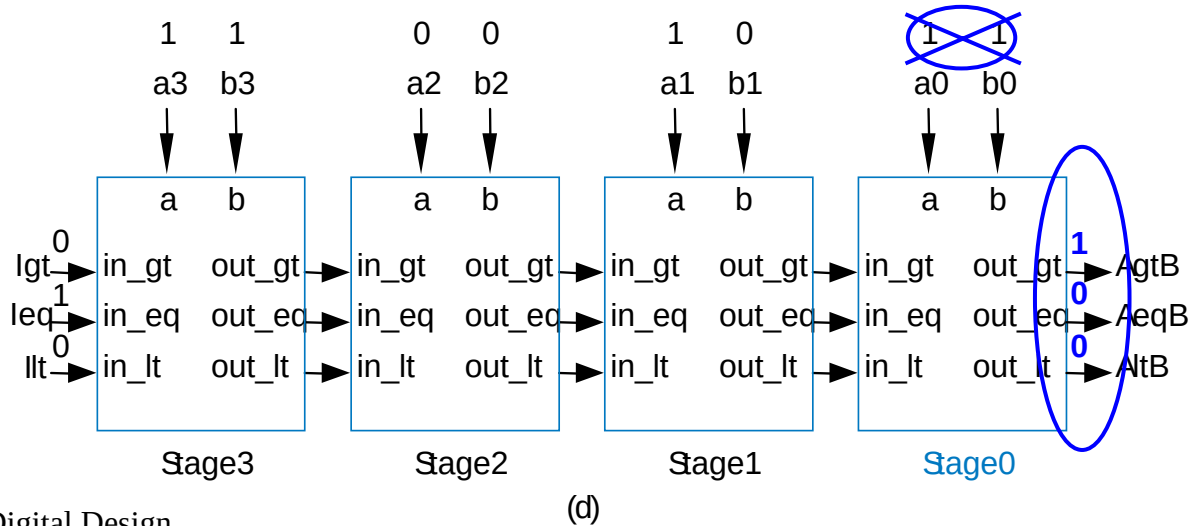
Ieq=1 causes this stage to compare

Magnitude Comparator

1011 = 1001 ?



(c)

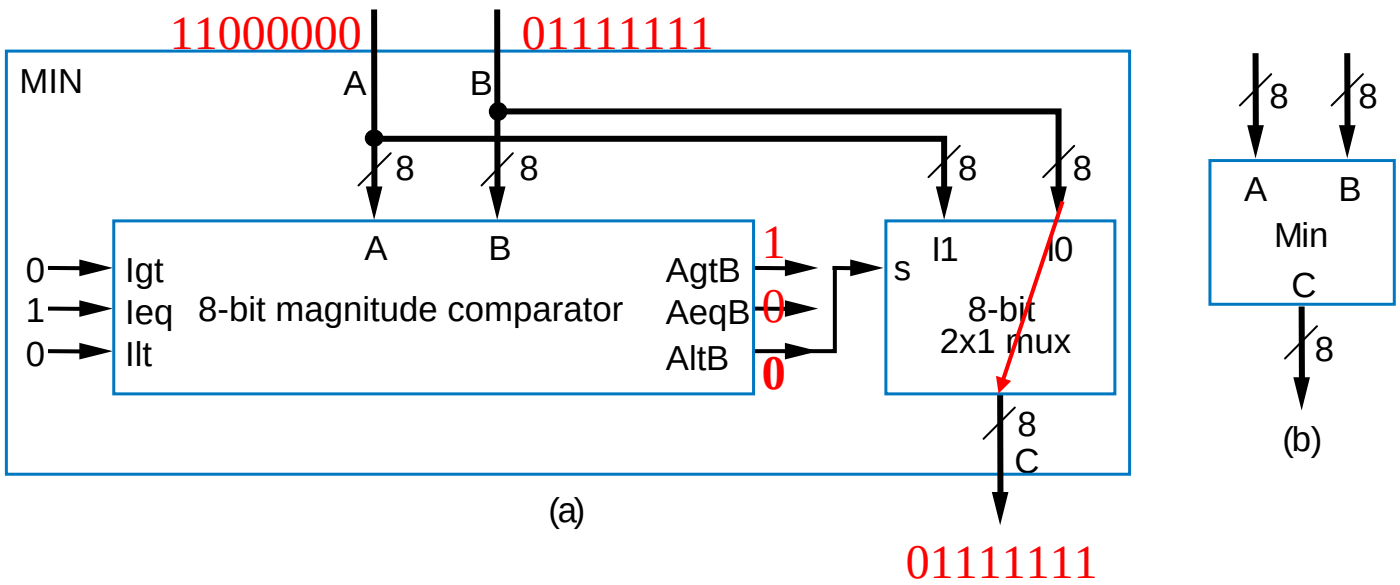


(d)

- Final answer appears on the right
- Takes time for answer to "ripple" from left to right
- Thus called "carry-ripple style" after the carry-ripple adder
  - Even though there's no "carry" involved

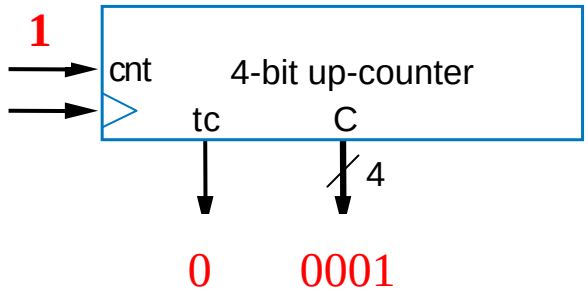
Magnitude Comparator Example: Minimum of Two Numbers

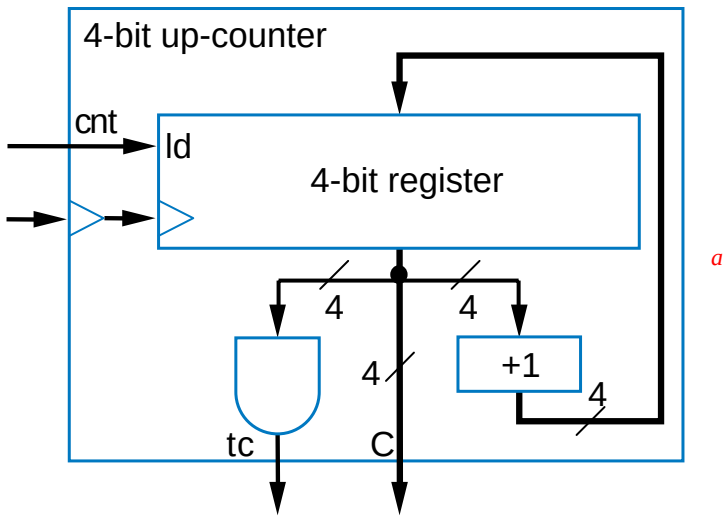
- Design a combinational component that computes the minimum of two 8-bit numbers
  - Solution: Use 8-bit magnitude comparator and 8-bit 2x1 mux
    - If A<B, pass B through mux. Else, pass A.



Counters

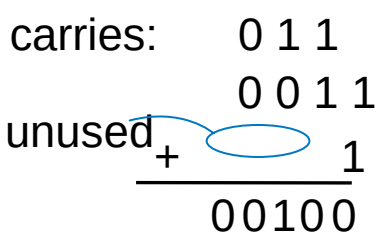
- ***N-bit up-counter***: N-bit register that can increment (add 1) to its own value on each clock cycle
  - 0000, 0001, 0010, 0011, ..., 1110, 1111, 0000
  - Note how count "rolls over" from 1111 to 0000
    - Terminal (last) count, tc, equals1 during value just before rollover
- Internal design
  - Register, incrementer, and N-input AND gate to detect terminal count

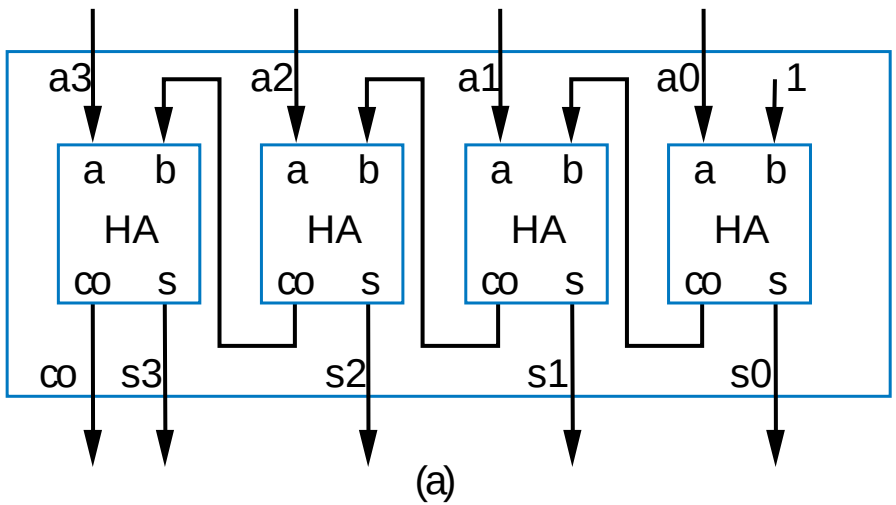


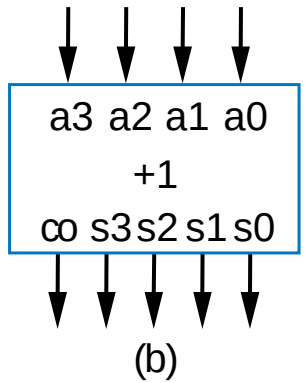


Incrementer

- Counter design used incrementer
- Incrementer design
  - Could use carry-ripple adder with B input set to 00...001
    - But when adding 00...001 to another number, the leading 0's obviously don't need to be considered -- so just two bits being added per column
  - Use half-adders (adds two bits) rather than full-adders (adds three bits)







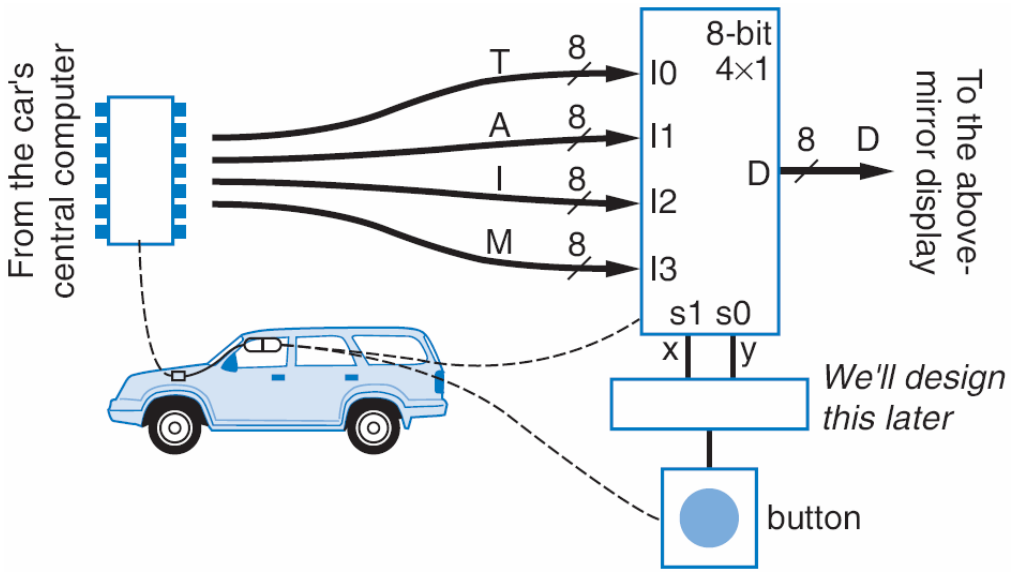
Incrementer

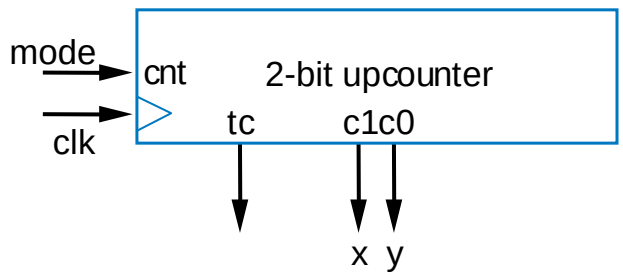
- Can build faster incrementer using combinational logic design process
  - Capture truth table
  - Derive equation for each output
    - c0 = a3a2a1a0
    - ...
    - s0 = a0'
  - Results in small and fast circuit
  - Note: works for small N -- larger N leads to exponential growth, like for N-bit adder

Inputs				Outputs				
a3	a2	a1	a0	c0	s3	s2	s1	s0
0	0	0	0	0	0	0	0	1
0	0	0	1	0	0	0	1	0
0	0	1	0	0	0	0	1	1
0	0	1	1	0	0	1	0	0
0	1	0	0	0	0	1	0	1
0	1	0	1	0	0	1	1	0
0	1	1	0	0	0	1	1	1
0	1	1	1	0	1	0	0	0
1	0	0	0	0	1	0	0	1
1	0	0	1	0	1	0	1	0
1	0	1	0	0	1	0	1	1
1	0	1	1	0	1	1	0	0
1	1	0	0	0	1	1	0	1
1	1	0	1	0	1	1	1	0
1	1	1	0	0	1	1	1	1
1	1	1	1	1	0	0	0	0

Counter Example: Mode in Above-Mirror Display

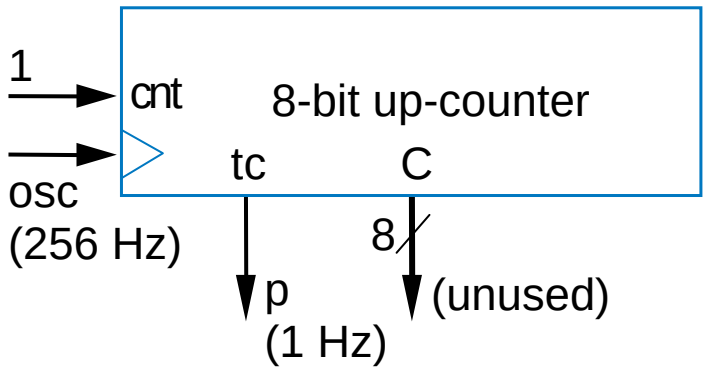
- Recall above-mirror display example from Chapter 2
  - Assumed component that incremented xy input each time button pressed: 00, 01, 10, 11, 00, 01, 10, 11, 00, ...
  - Can use 2-bit up-counter
    - Assumes mode=1 for just one clock cycle during each button press
      - Recall "Button press synchronizer" example from Chapter 3





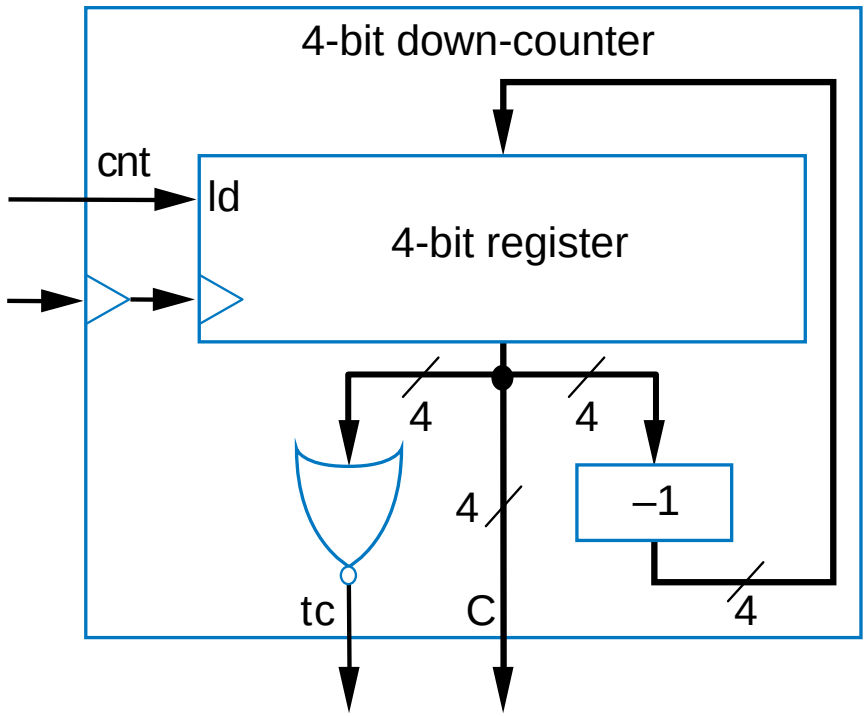
Counter Example: 1 Hz Pulse Generator Using 256 Hz Oscillator

- Suppose have 256 Hz oscillator, but want 1 Hz pulse
  - 1 Hz is 1 pulse per second -- useful for keeping time
  - Design using 8-bit up-counter, use tc output as pulse
    - Counts from 0 to 255 (256 counts), so pulses tc every 256 cycles



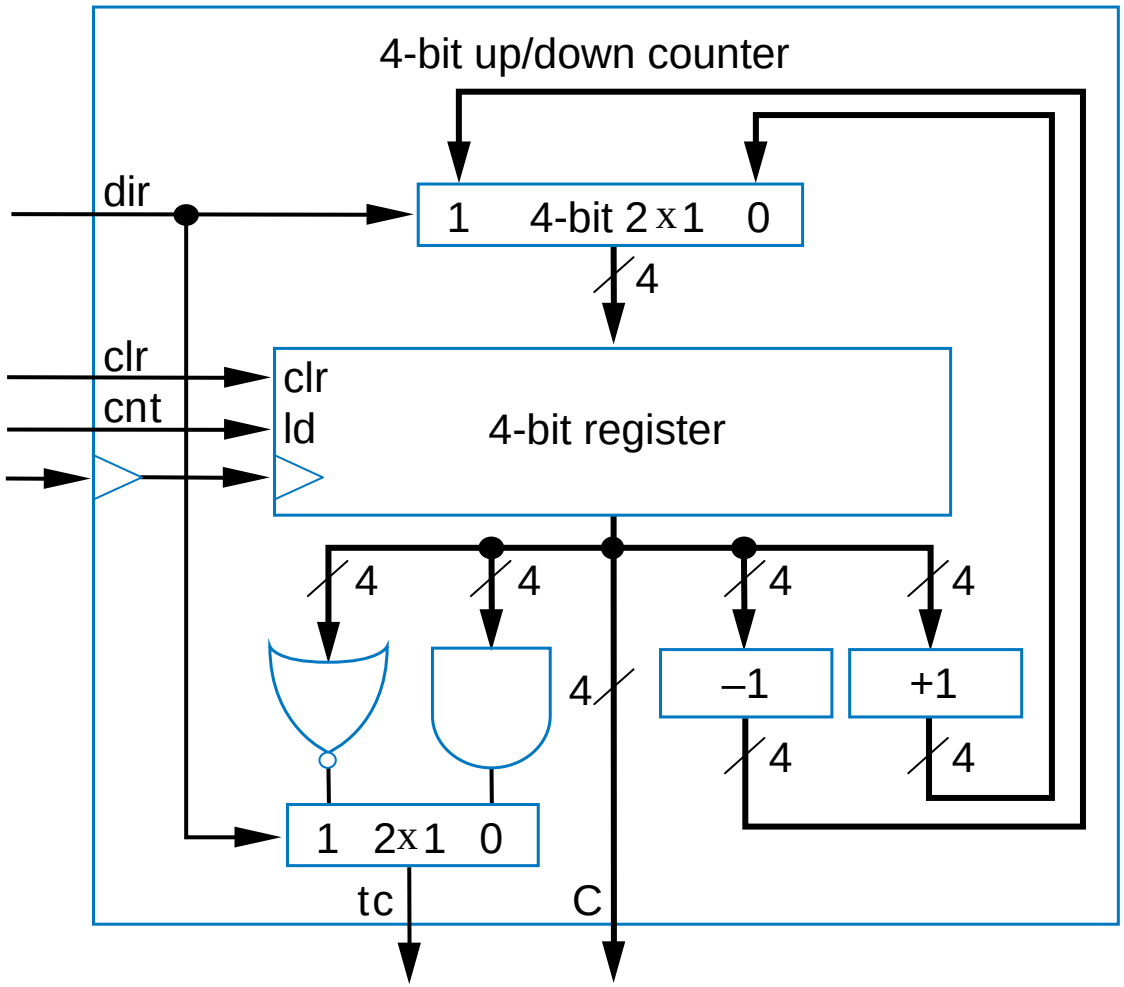
Down-Counter

- 4-bit down-counter
  - 1111, 1110, 1101, 1100, …, 0011, 0010, 0001, 0000, 1111, …
  - Terminal count is 0000
    - Use NOR gate to detect
  - Need decrementer (-1) – design like designed incrementer



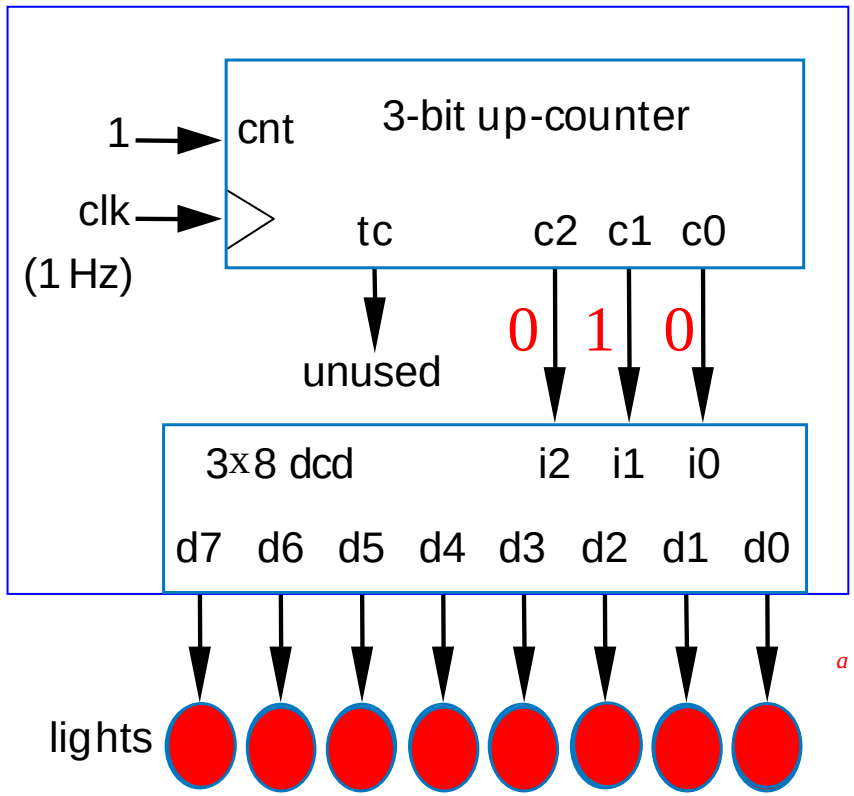
Up/Down-Counter

- Can count either up or down
  - Includes both incrementer and decrementer
  - Use dir input to select, using 2x1: dir=0 means up
  - Likewise, dir selects appropriate terminal count value



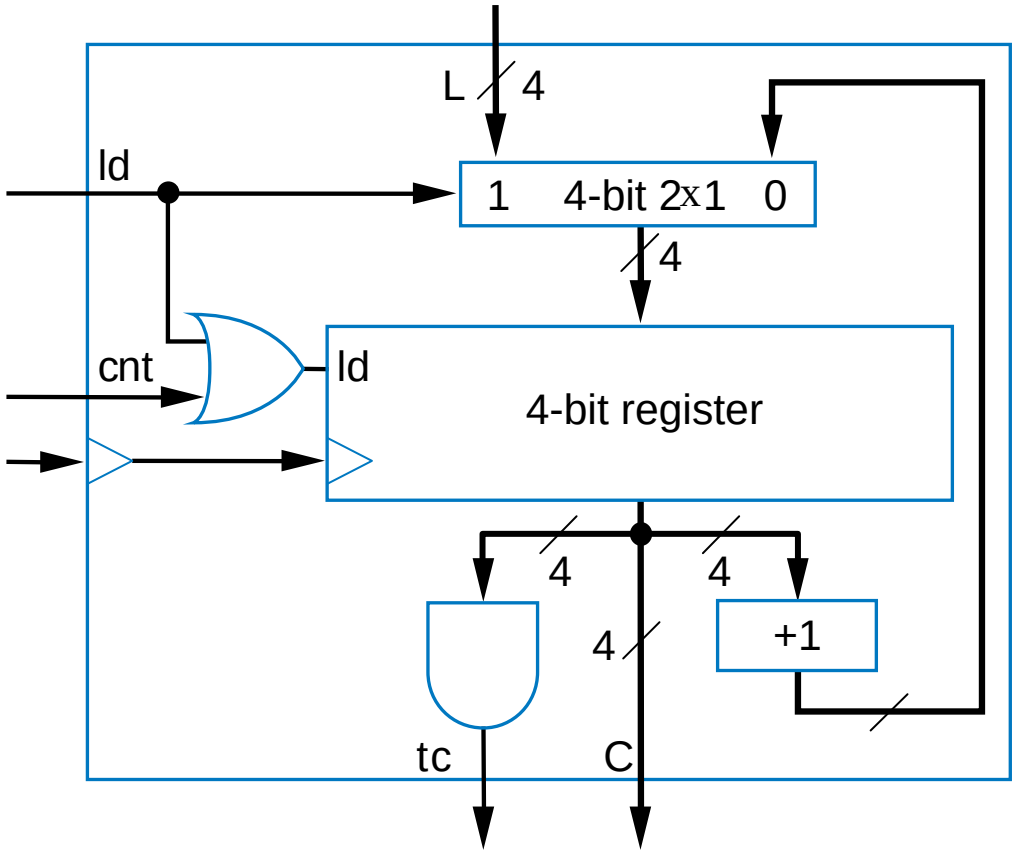
Counter Example: Light Sequencer

- Illuminate 8 lights from right to left, one at a time, one per second
- Use 3-bit up-counter to counter from 0 to 7
- Use 3x8 decoder to illuminate appropriate light
- Note: Used ***3-bit*** counter with 3x8 decoder
  - *NOT* an *8-bit* counter – why not?



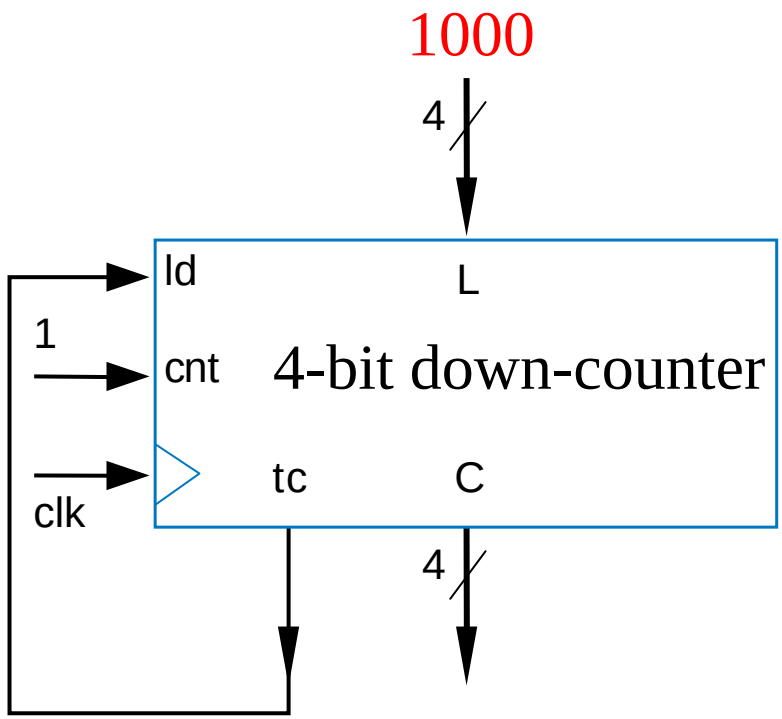
Counter with Parallel Load

- Up-counter that can be loaded with external value
  - Designed using 2x1 mux – ld input selects incremented value or external value
  - Load the internal register when loading external value or when counting



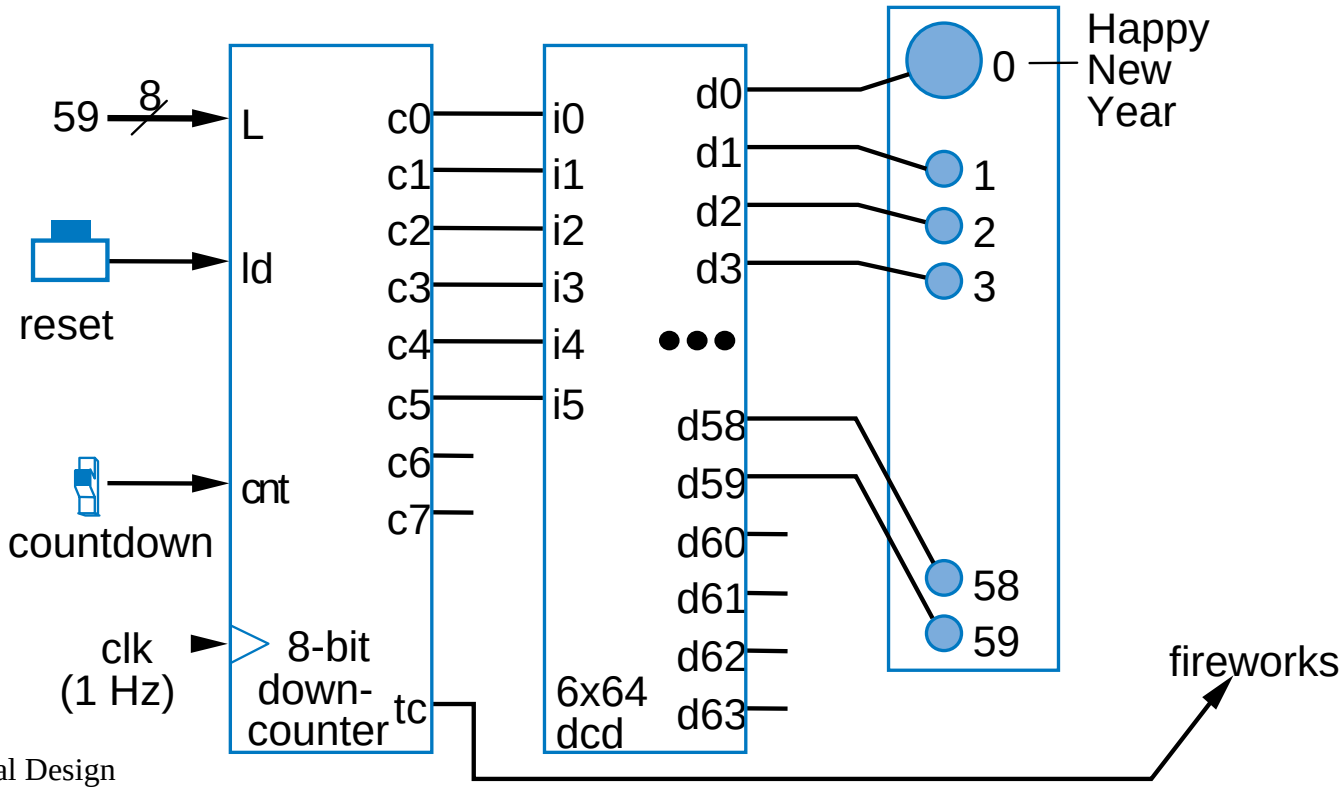
Counter with Parallel Load

- Useful to create pulses at specific multiples of clock
  - Not just at N-bit counter's natural wrap-around of 2^N
- Example: Pulse every 9 clock cycles
  - Use 4-bit down-counter with parallel load
  - Set parallel load input to 8 (1000)
  - Use terminal count to reload
    - When count reaches 0, next cycle loads 8.
  - Why load 8 and not 9? Because 0 is included in count sequence:
    - 8, 7, 6, 5, 4, 3, 2, 1, 0 → 9 counts



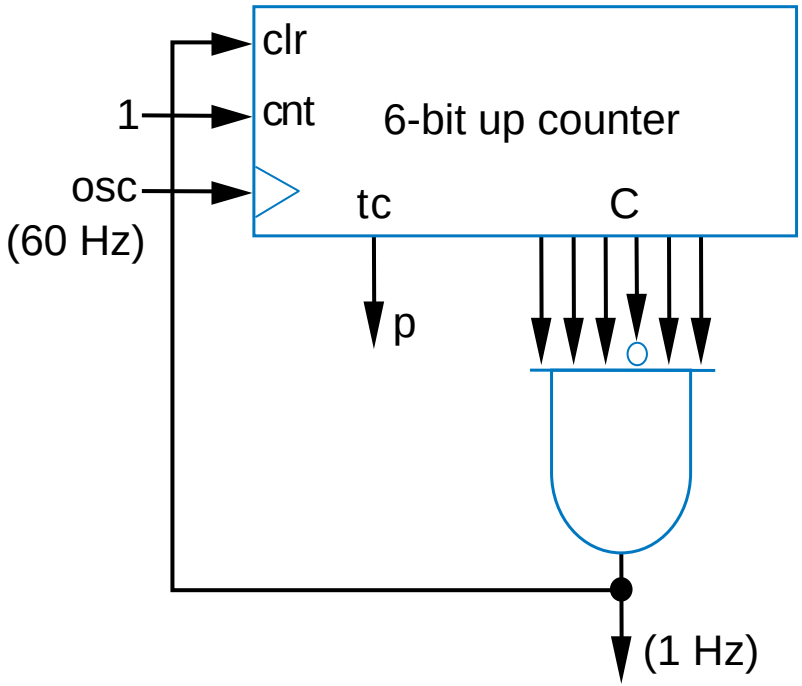
Counter Example:
New Year's Eve Countdown Display

- Chapter 2 example previously used microprocessor to counter from 59 down to 0 in binary
- Can use 8-bit (or 7- or 6-bit) down-counter instead, initially loaded with 59



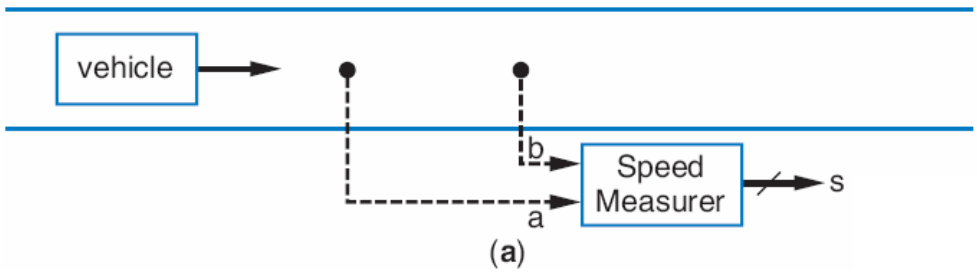
Counter Example:
1 Hz Pulse Generator from 60 Hz Clock

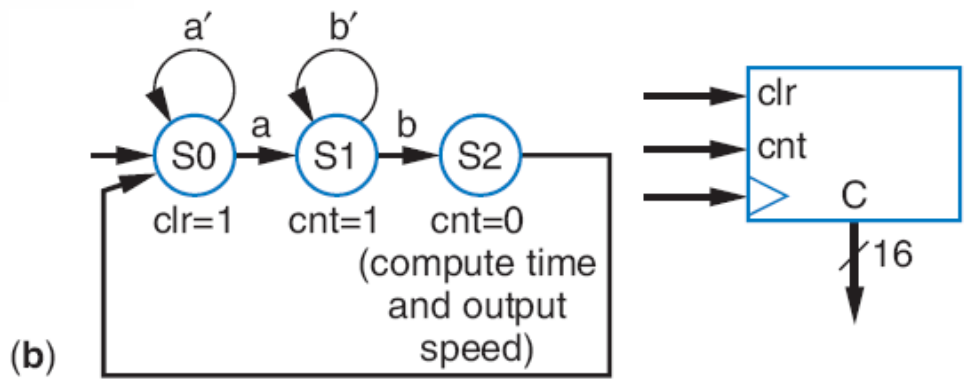
- U.S. electricity standard uses 60 Hz signal
  - Device may convert that to 1 Hz signal to count seconds
- Use 6-bit up-counter
  - Can count from 0 to 63
  - Create simple logic to detect 59 (for 60 counts)
    - Use to clear the counter back to 0 (or to load 0)



Timer

- A type of counter used to measure time
  - If we know the counter's clock frequency and the count, we know the time that's been counted
- Example: Compute car's speed using two sensors
  - First sensor (a) clears and starts timer
  - Second sensor (b) stops timer
  - Assuming clock of 1kHz, timer output represents time to travel between sensors. Knowing the distance, we can compute speed





Multiplier – Array Style

- Can build multiplier that mimics multiplication by hand
  - Notice that multiplying multiplicand by 1 is same as ANDing with 1

```
  0110   (the top number is called the multiplicand)
  0011   (the bottom number is called the multiplier)
  ----   (each row below is called a partial product)
  0110   (because the rightmost bit of the multiplier is 1, and 0110*1=0110)
 0110    (because the second bit of the multiplier is 1, and 0110*1=0110)
0000     (because the third bit of the multiplier is 0, and 0110*0=0000)
+0000    (because the leftmost bit of the multiplier is 0, and 0110*0=0000)
--------
00010010 (the product is the sum of all the partial products: 18, which is 6*3)
```

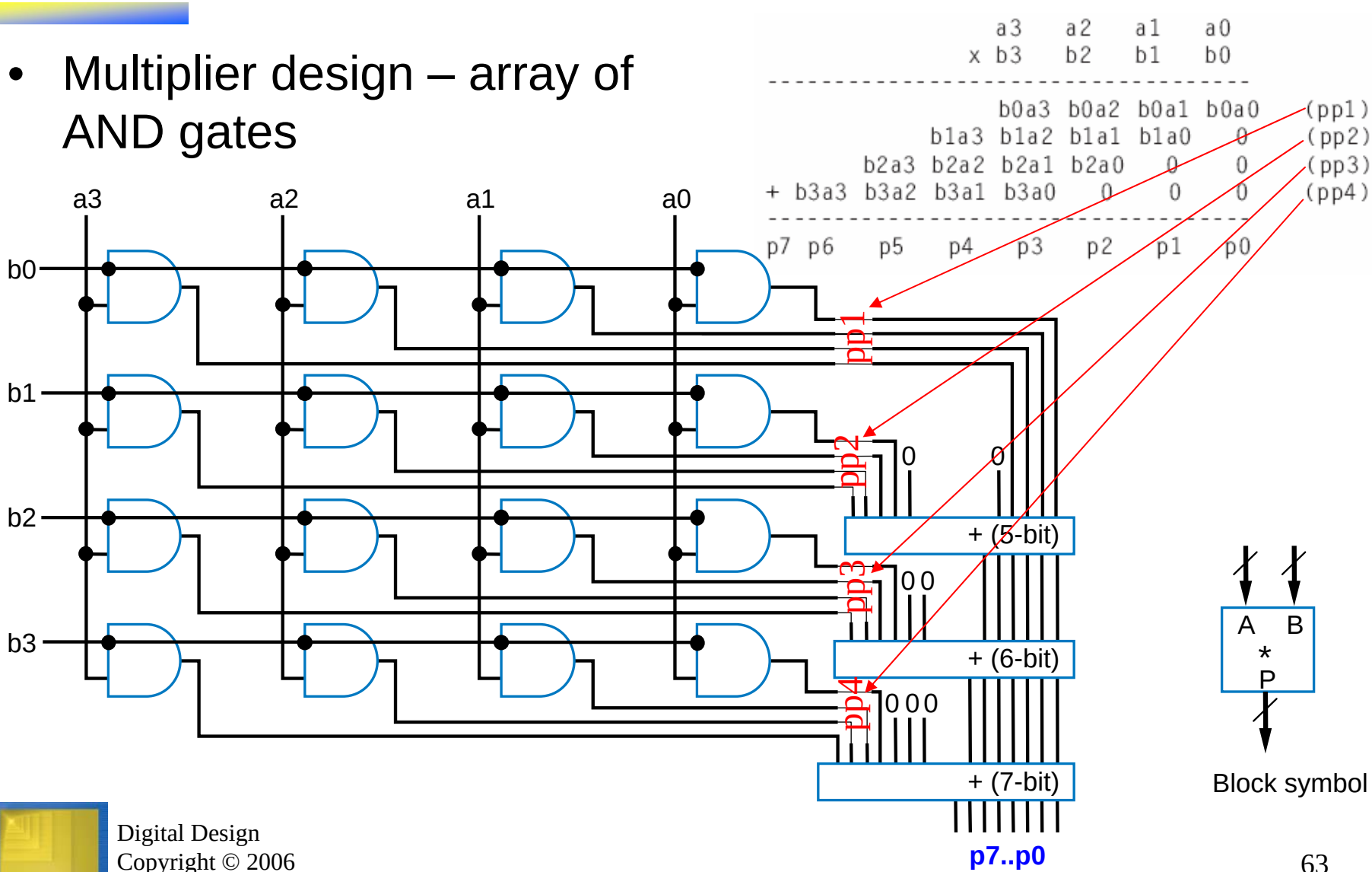
Multiplier – Array Style

- Generalized representation of multiplication by hand

```
                    a3    a2    a1    a0
                  x b3    b2    b1    b0
-------------------------------------------
                    b0a3  b0a2  b0a1  b0a0    (pp1)
              b1a3  b1a2  b1a1  b1a0   0     (pp2)
        b2a3  b2a2  b2a1  b2a0   0     0     (pp3)
+ b3a3  b3a2  b3a1  b3a0   0     0     0     (pp4)
-------------------------------------------
p7  p6    p5    p4    p3    p2    p1    p0
```

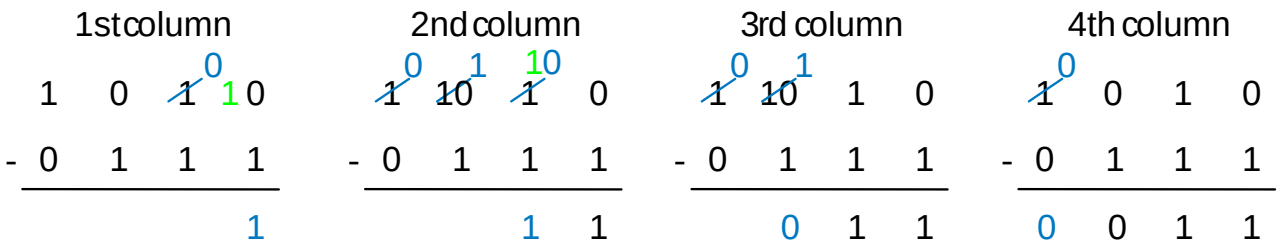
Multiplier – Array Style

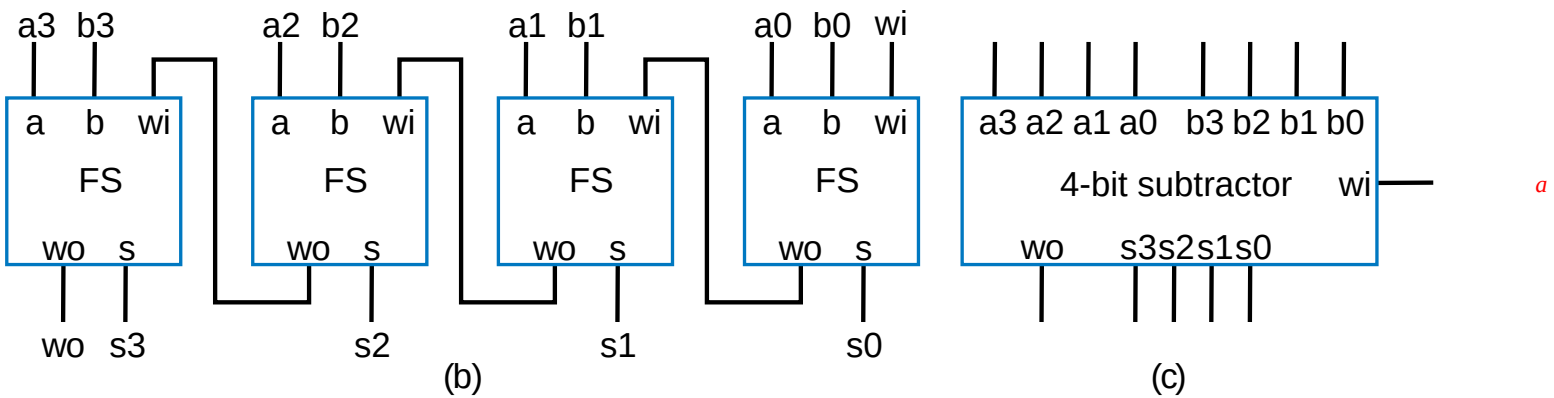
- Multiplier design – array of AND gates



Subtractor

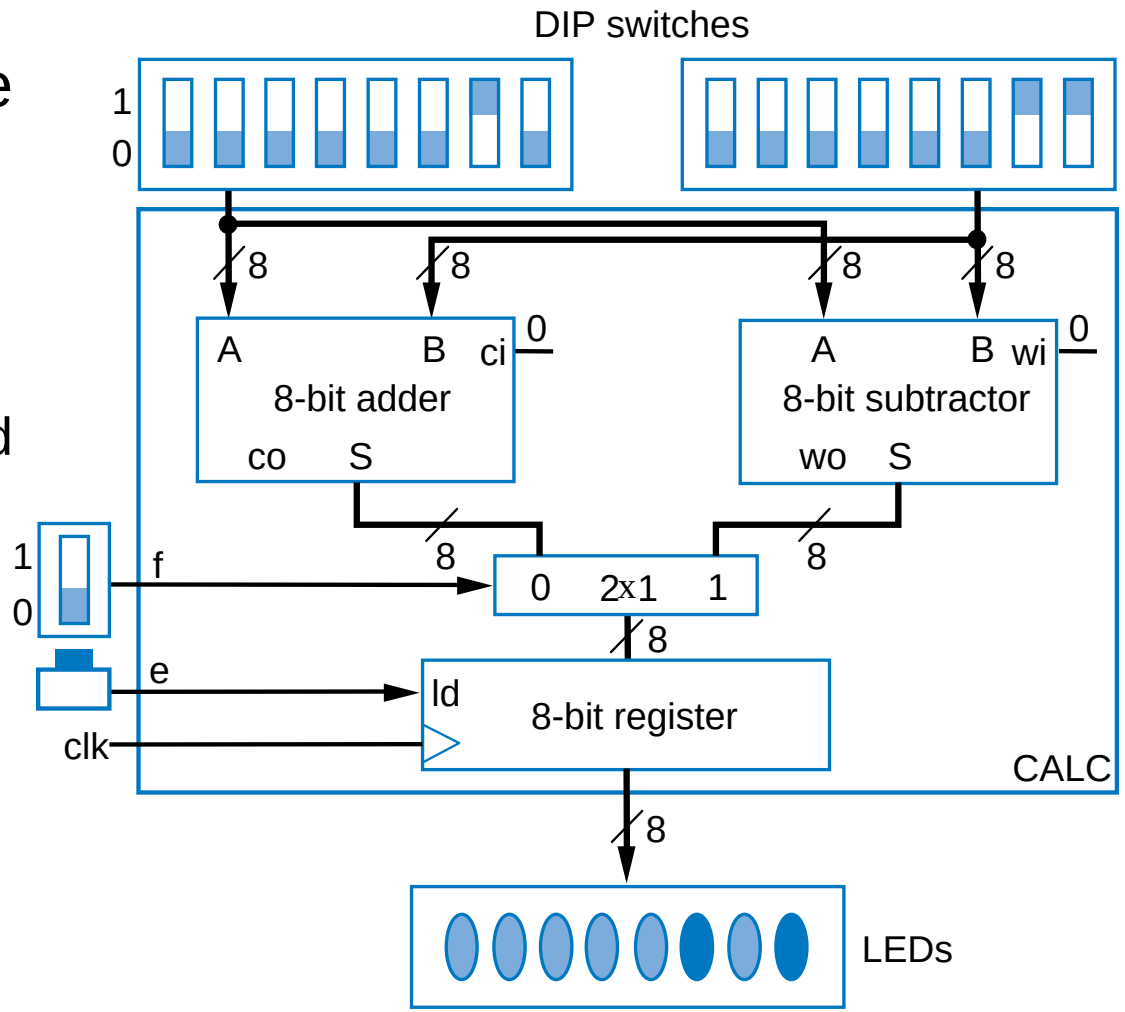
- Can build subtractor as we built carry-ripple adder
  - Mimic subtraction by hand
  - Compute borrows from columns on left
    - Use full-subtractor component:
      - wi is borrow by column on right, wo borrow from column on left





Subtractor Example: DIP-Switch Based Adding/Subtracting Calculator

- Extend earlier calculator example
  - Switch f indicates whether want to add (f=0) or subtract (f=1)
  - Use subtractor and 2x1 mux



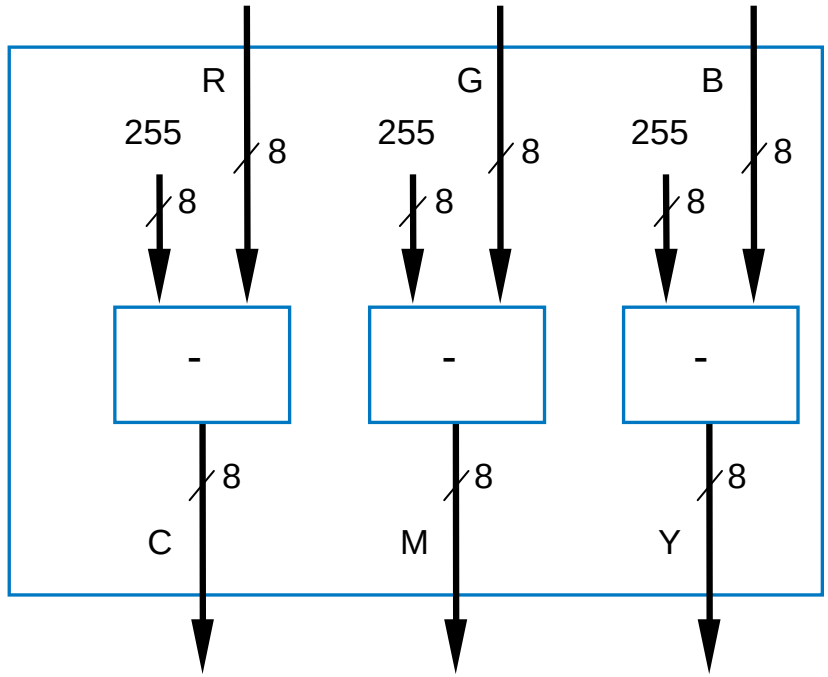
Subtractor Example:
Color Space Converter – RGB to CMYK

- Color
  - Often represented as weights of three colors: red, green, and blue (RGB)
    - Perhaps 8 bits each, so specific color is 24 bits
      - White: R=11111111, G=11111111, B=11111111
      - Black: R=00000000, G=00000000, B=00000000
      - Other colors: values in between, e.g., R=00111111, G=00000000, B=00001111 would be a reddish purple
  - Good for computer monitors, which mix red, green, and blue lights to form all colors

- Printers use opposite color scheme
  - Because inks *absorb* light
  - Use complementary colors of RGB: Cyan (absorbs red), reflects green and blue, Magenta (absorbs green), and Yellow (absorbs blue)

Subtractor Example:
Color Space Converter – RGB to CMYK

- Printers must quickly convert RGB to CMY
  - C=255-R, M=255-G, Y=255-B
  - Use subtractors as shown

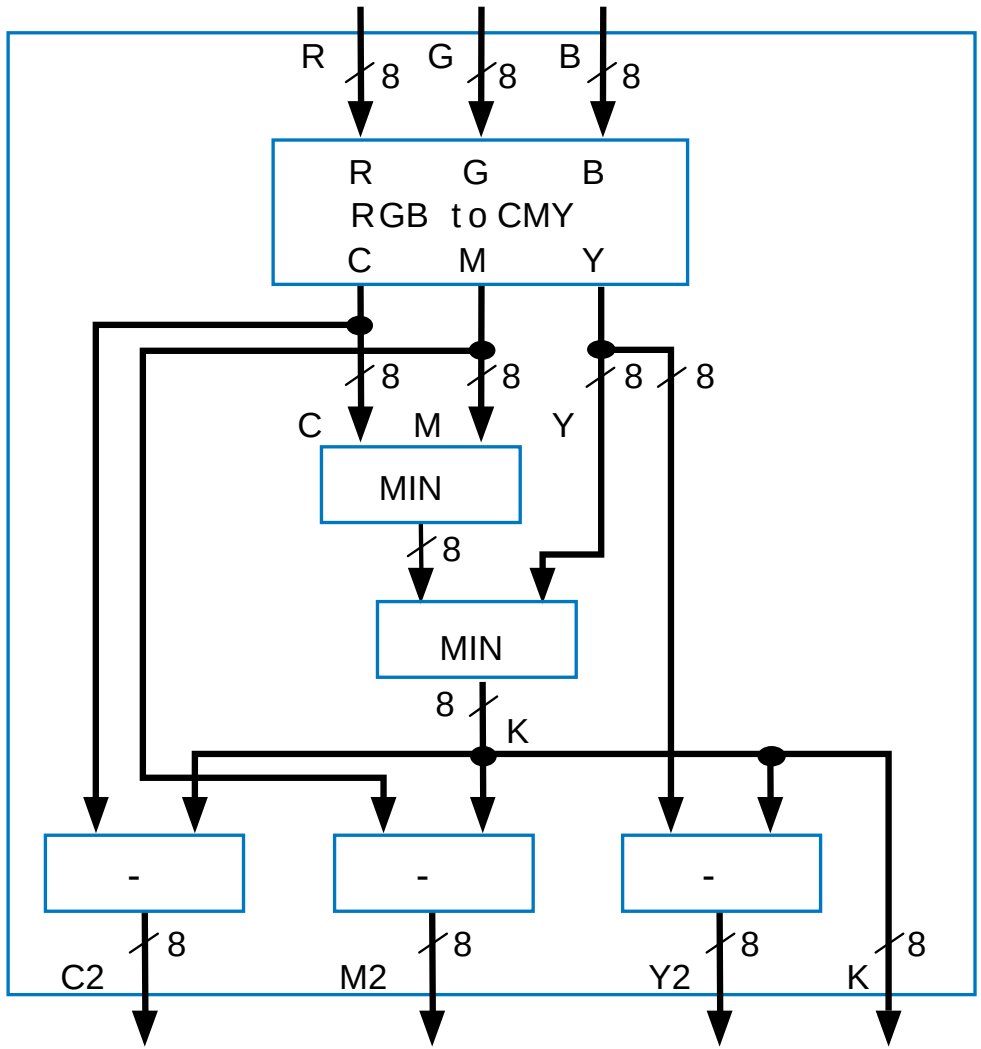


Subtractor Example:
Color Space Converter – RGB to CMYK

- Try to save colored inks
  - Expensive
  - Imperfect – mixing C, M, Y doesn't yield good-looking black
- Solution: Factor out the black or gray from the color, print that part using black ink
  - e.g., CMY of (250,200,200)= (200,200,200) + (50,0,0).
    - (200,200,200) is a dark gray – use black ink

Subtractor Example: Color Space Converter – RGB to CMYK

- Call black part K
  - (200,200,200): K=200
  - (Letter “B” already used for blue)
- Compute minimum of C, M, Y values
  - Use MIN component designed earlier, using comparator and mux, to compute K
  - Output resulting K value, and subtract K value from C, M, and Y values
  - Ex: Input of (250,200,200) yields output of (50,0,0,200)



Representing Negative Numbers: Two's Complement

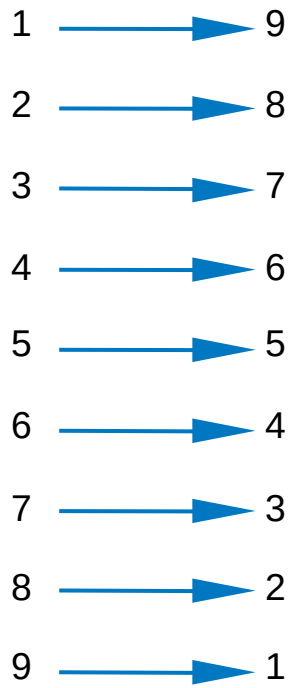
- Negative numbers common
  - How represent in binary?
- Signed-magnitude
  - Use leftmost bit for sign bit
    - So -5 would be:

      1101 using four bits

      10000101 using eight bits
- Better way: Two's complement
  - Big advantage: Allows us to perform subtraction using addition
  - Thus, only need adder component, no need for separate subtractor component!

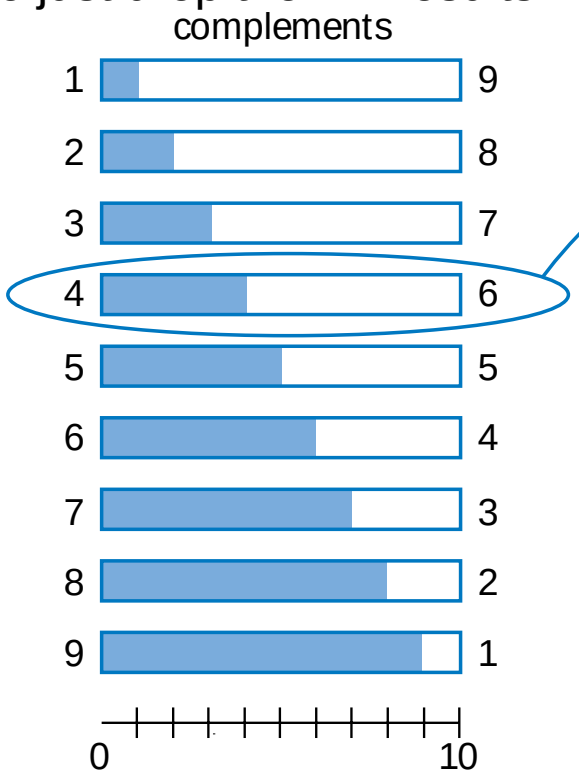
Ten's Complement

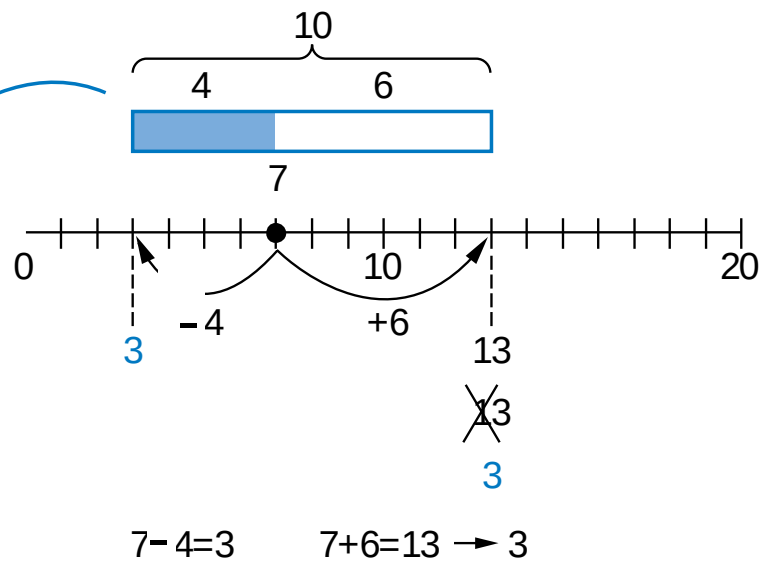
- Before introducing two's complement, let's consider ten's complement
  - But, be aware that computers DO NOT USE TEN'S COMPLEMENT. Introduced for intuition only.
  - Complements for each base ten number shown to right – Complement is the number that when added results in 10



Ten's Complement

- Nice feature of ten's complement
  - Instead of subtracting a number, adding its complement results in answer exactly 10 too much
  - So just drop the 1 – results in subtracting using addition only





Adding the complement results in an answer exactly 10 too much – dropping the tens column gives the right answer.

Two's Complement is Easy to Compute:
Just Invert Bits and Add 1

- Hold on!
  - Sure, adding the ten's complement achieves subtraction using addition only
  - But don't we have to perform *subtraction* to have determined the complement in the first place? e.g., we only know that the complement of 4 is 6 by subtracting 10-4=6 in the first place.
- True – but in binary, it turns out that the two's complement can be computed **easily**
  - Two's complement of 011 is 101, because 011 + 101 is 1000
  - Could compute complement of 011 as 1000 – 011 = 101
  - **Easier method: Just invert all the bits, and add 1**
  - The complement of 011 is 100+1 = 101 -- it works!

Q: What is the two's complement of 0101?

A: 1010+1=1011

(check: 0101+1011=10000)

Q: What is the two's complement of 0011?

A: 1100+1=1101

Two's Complement Subtractor Built with an Adder

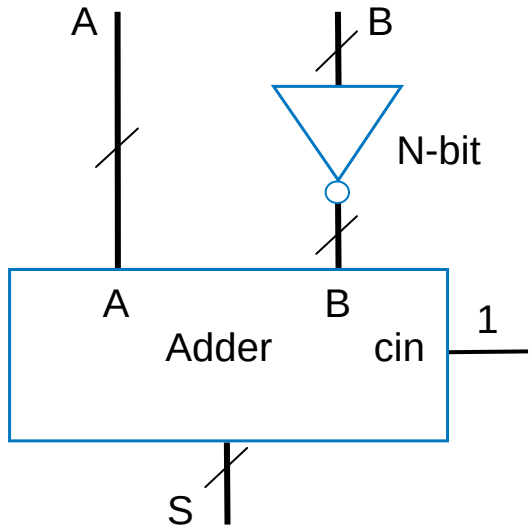
- Using two's complement

  A – B = A + (-B)

  = A + (two's complement of B)

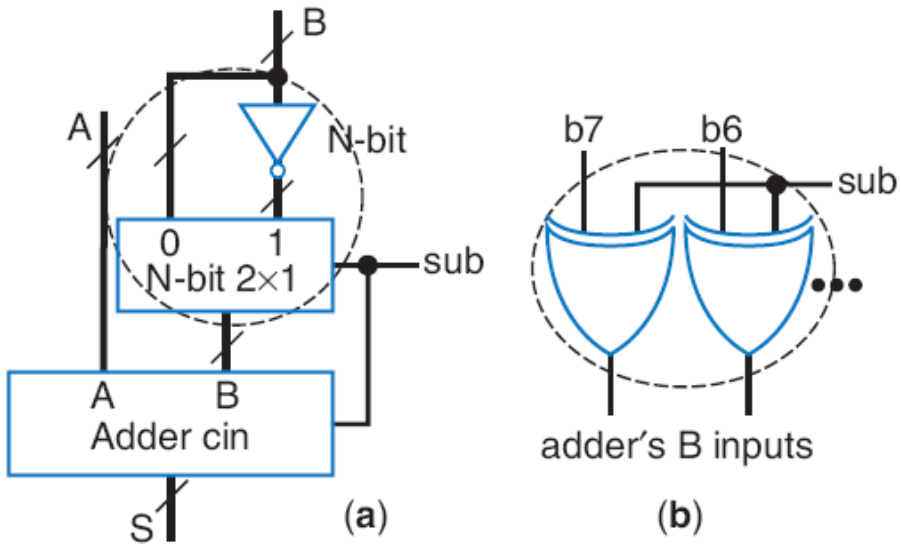
  = A + invert_bits(B) + 1

- So build subtractor using adder by inverting B's bits, and setting carry in to 1



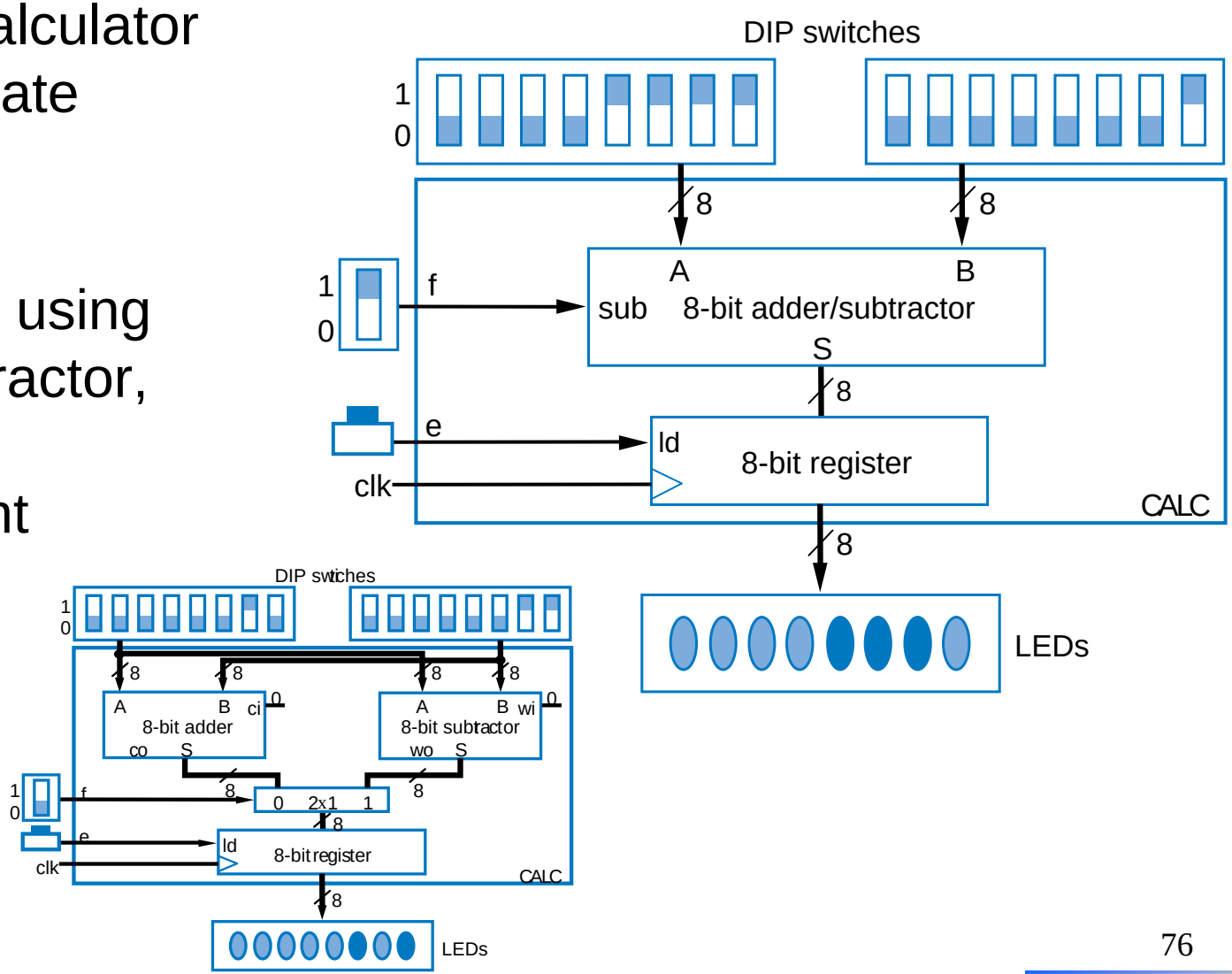
Adder/Subtractor

- Adder/subtractor: control input determines whether add or subtract
  - Can use 2x1 mux – sub input passes either B or inverted B
  - Alternatively, can use XOR gates – if sub input is 0, B's bits pass through; if sub input is 1, XORs invert B's bits



Adder/Subtractor Example: Calculator

- Previous calculator used separate adder and subtractor
- Improve by using adder/subtractor, and two's complement numbers

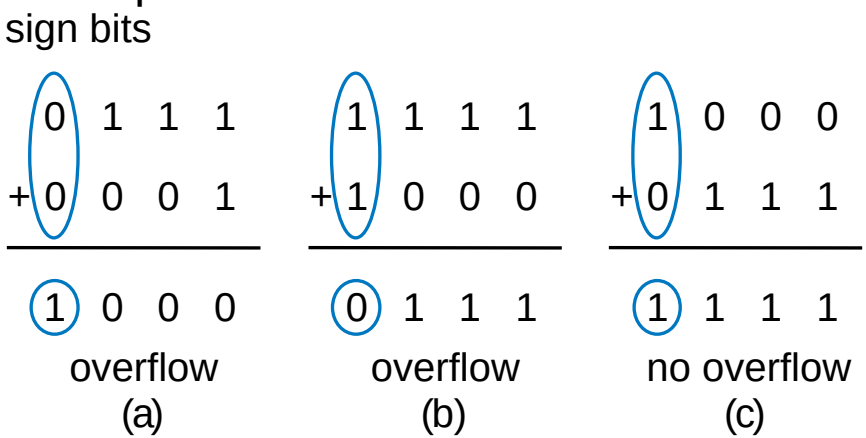


Overflow

- Sometimes result can't be represented with given number of bits
  - Either too large magnitude of positive or negative
  - e.g., 4-bit two's complement addition of 0111+0001 (7+1=8). But 4-bit two's complement can't represent number >7
    - 0111+0001 = 1000 WRONG answer, 1000 in two's complement is -8, not +8
  - Adder/subtractor should indicate when overflow has occurred, so result can be discarded

Detecting Overflow: Method 1

- Assuming 4-bit two's complement numbers, can detect overflow by detecting when the two numbers' sign bits are the same but are different from the result's sign bit
  - If the two numbers' sign bits are different, overflow is impossible
    - Adding a positive and negative can't exceed largest magnitude positive or negative
- Simple circuit
  - overflow = a3'b3's3 + a3b3s3'
  - Include "overflow" output bit on adder/subtractor



If the numbers' sign bits have the same value, which differs from the result's sign bit, overflow has occurred.

Detecting Overflow: Method 2

- Even simpler method: Detect difference between carry-in to sign bit and carry-out from sign bit
- Yields simpler circuit: overflow = c3 xor c4

```
  1  1  1          0  0  0          0  0  0
  0  1  1  1       1  1  1  1       1  0  0  0
+ 0  0  0  1     + 1  0  0  0     + 0  1  1  1
-------------    -------------    -------------
0 1  0  0  0     1 0  1  1  1     0 1  1  1  1
```

overflow (a) | overflow (b) | no overflow (c)

If the carry into the sign bit column differs from the carry out of that column, overflow has occurred.

Arithmetic-Logic Unit: ALU

- ***ALU***: Component that can perform any of various arithmetic (add, subtract, increment, etc.) and logic (AND, OR, etc.) operations, based on control inputs
- Motivation:
  - Suppose want multi-function calculator that not only adds and subtracts, but also increments, ANDs, ORs, XORs, etc.

TABLE 4.2 Desired calculator operations

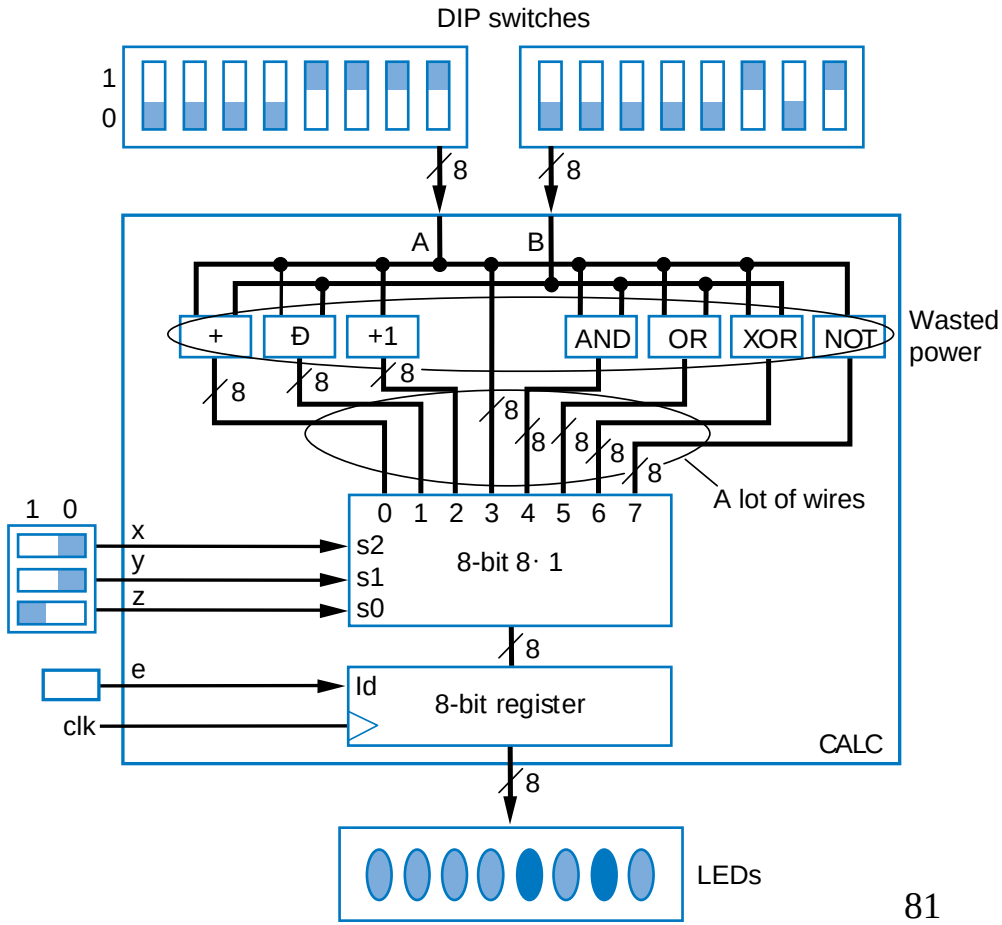
Inputs			Operation	Sample output if A=00001111, B=00000101
x	y	z		
0	0	0	S = A + B	S=00010100
0	0	1	S = A - B	S=00001010
0	1	0	S = A + 1	S=00010000
0	1	1	S = A	S=00001111
1	0	0	S = A AND B (bitwise AND)	S=00000101
1	0	1	S = A OR B (bitwise OR)	S=00001111
1	1	0	S = A XOR B (bitwise XOR)	S=00001010
1	1	1	S = NOT A (bitwise complement)	S=11110000

Multifunction Calculator without an ALU

- Can build multifunction calculator using separate components for each operation, and muxes
  - But too many wires, and wasted power computing all those operations when at any time you only use

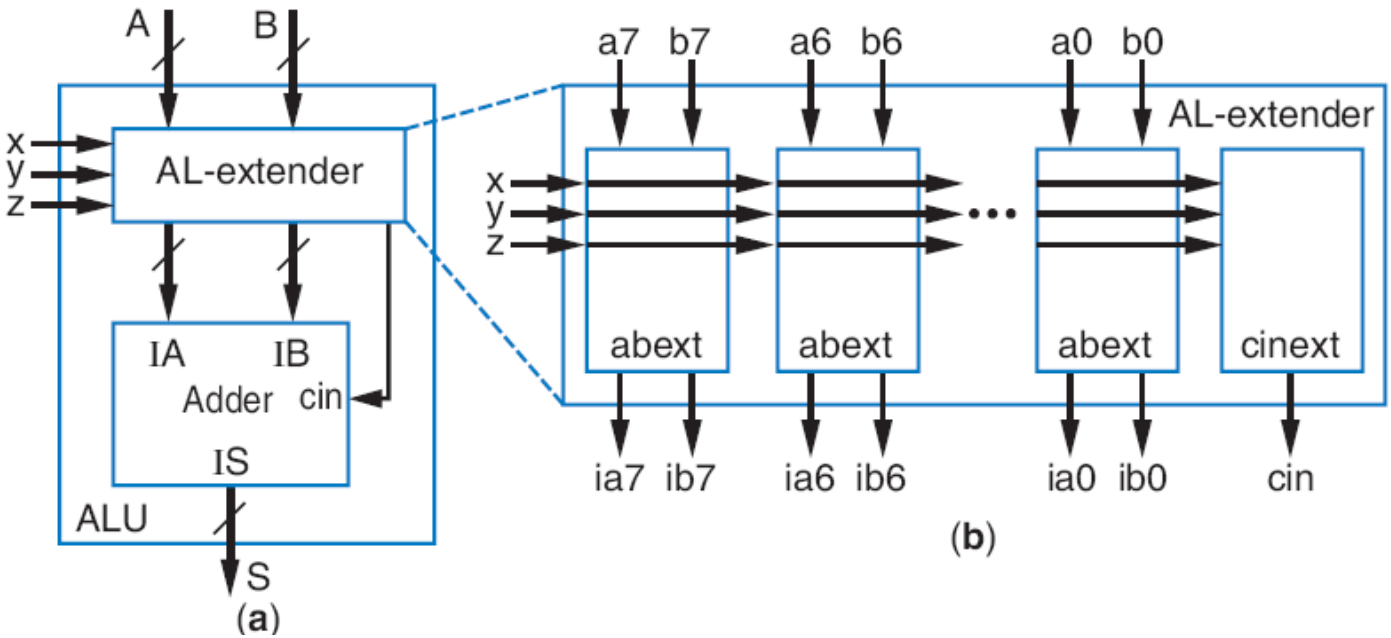
TABLE 4.2 Desired calculator operations

Inputs			Operation	Sample output if A=00001111, B=00000101
x	y	z		
0	0	0	S = A + B	S=00010100
0	0	1	S = A - B	S=00001010
0	1	0	S = A + 1	S=00010000
0	1	1	S = A	S=00001111
1	0	0	S = A AND B (bitwise AND)	S=00000101
1	0	1	S = A OR B (bitwise OR)	S=00001111
1	1	0	S = A XOR B (bitwise XOR)	S=00001010
1	1	1	S = NOT A (bitwise complement)	S=11110000



ALU

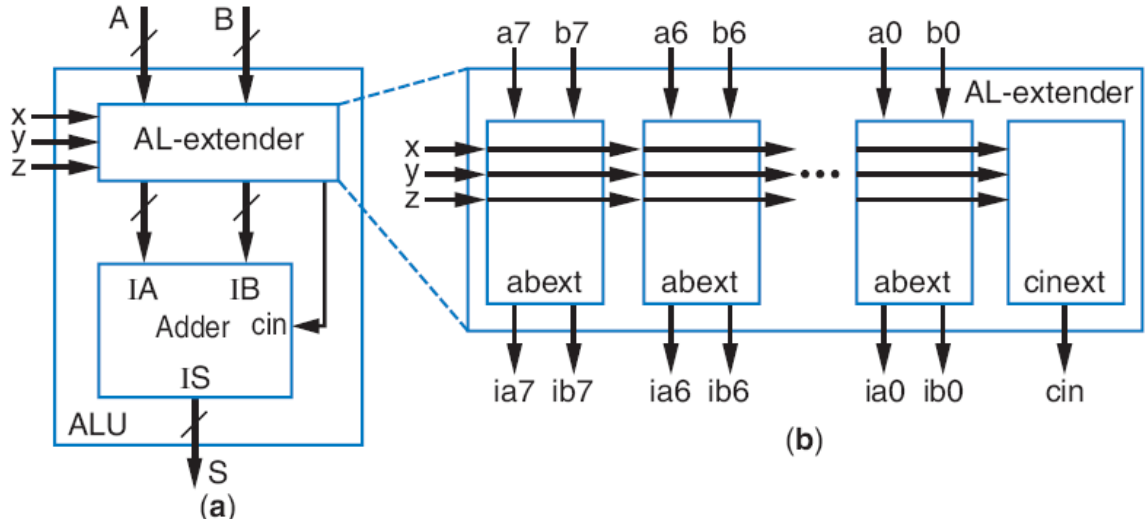
- More efficient design uses ALU
  - ALU design not just separate components multiplexed (same problem as previous slide!),
  - Instead, ALU design uses single adder, plus logic in front of adder's A and B inputs
    - Logic in front is called an arithmetic-logic extender
  - Extender modifies the A and B inputs such that desired operation will appear at output of the adder



Arithmetic-Logic Extender in Front of ALU

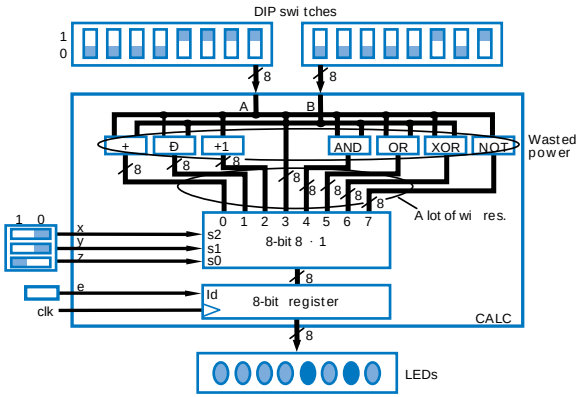
TABLE 4.2 Desired calculator operations

Inputs x	y	z	Operation	Sample output if A=00001111, B=00000101
0	0	0	S = A + B	S=00010100
0	0	1	S = A - B	S=00001010
0	1	0	S = A + 1	S=00010000
0	1	1	S = A	S=00001111
1	0	0	S = A AND B (bitwise AND)	S=00000101
1	0	1	S = A OR B (bitwise OR)	S=00001111
1	1	0	S = A XOR B (bitwise XOR)	S=00001010
1	1	1	S = NOT A (bitwise complement)	S=11110000

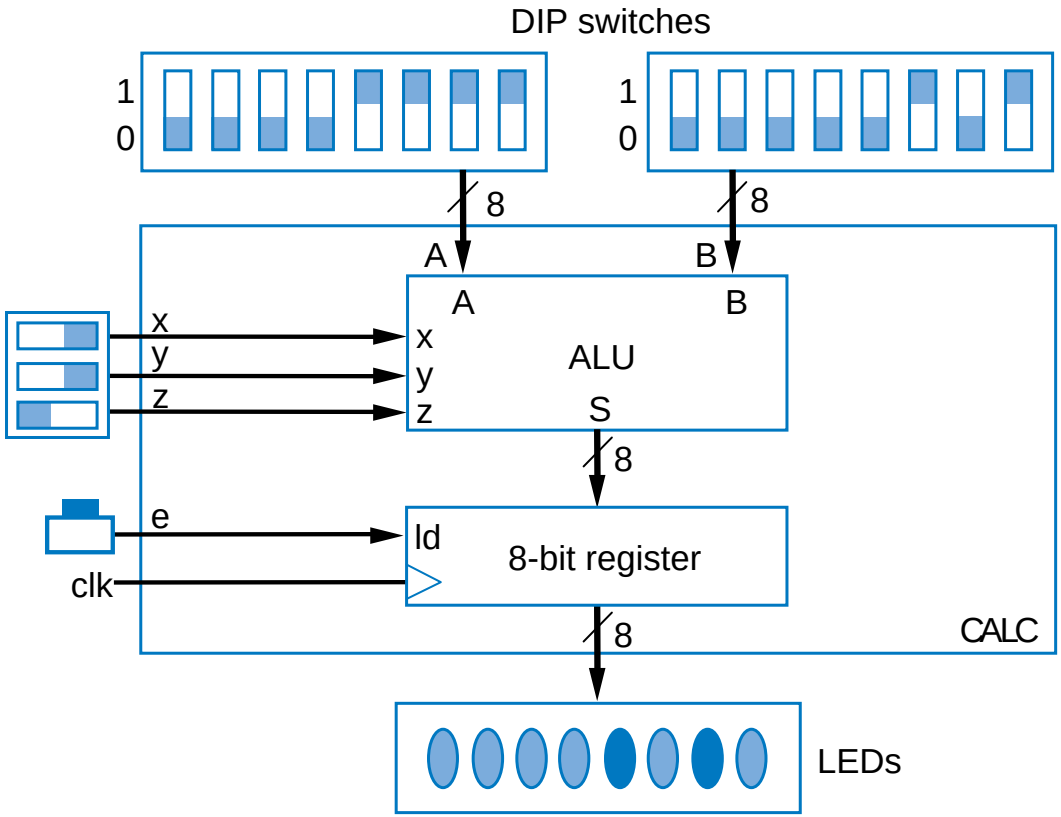


- xyz=000: Want S=A+B – just pass a to ia, b to ib, and set cin=0
- xyz=001: Want S=A-B – pass a to ia, b' to ib, and set cin=1
- xyz=010: Want S=A+1 – pass a to ia, set ib=0, and set cin=1
- xyz=011: Want S=A – pass a to ia, set ib=0, and set cin=0
- xyz=1000: Want S=A AND B – set ia=a*b, b=0, and cin=0
- others: likewise
- Based on above, create logic for ia(x,y,z,a,b) and ib(x,y,z,a,b) for each abext, and create logic for cin(x,y,z), to complete design of the AL-extender component

ALU Example: Multifunction Calculator

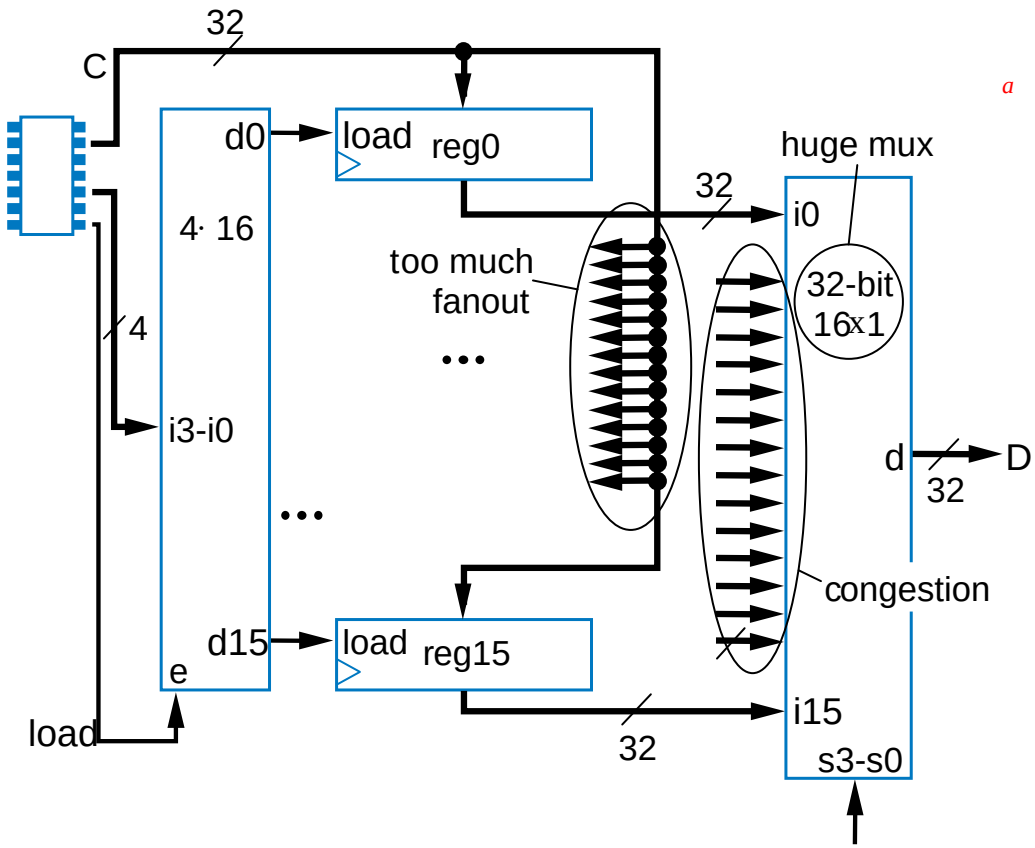


- Design using ALU is elegant and efficient
  - No mass of wires
  - No big waste of power



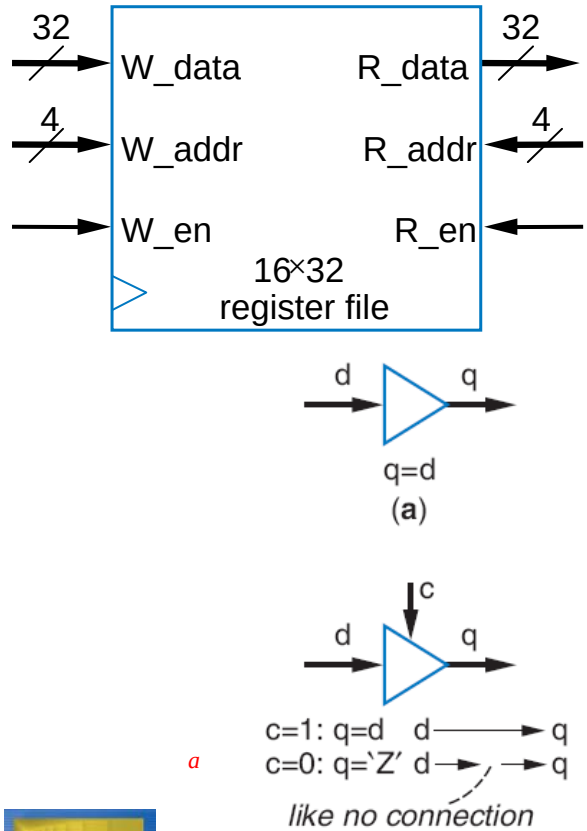
Register Files

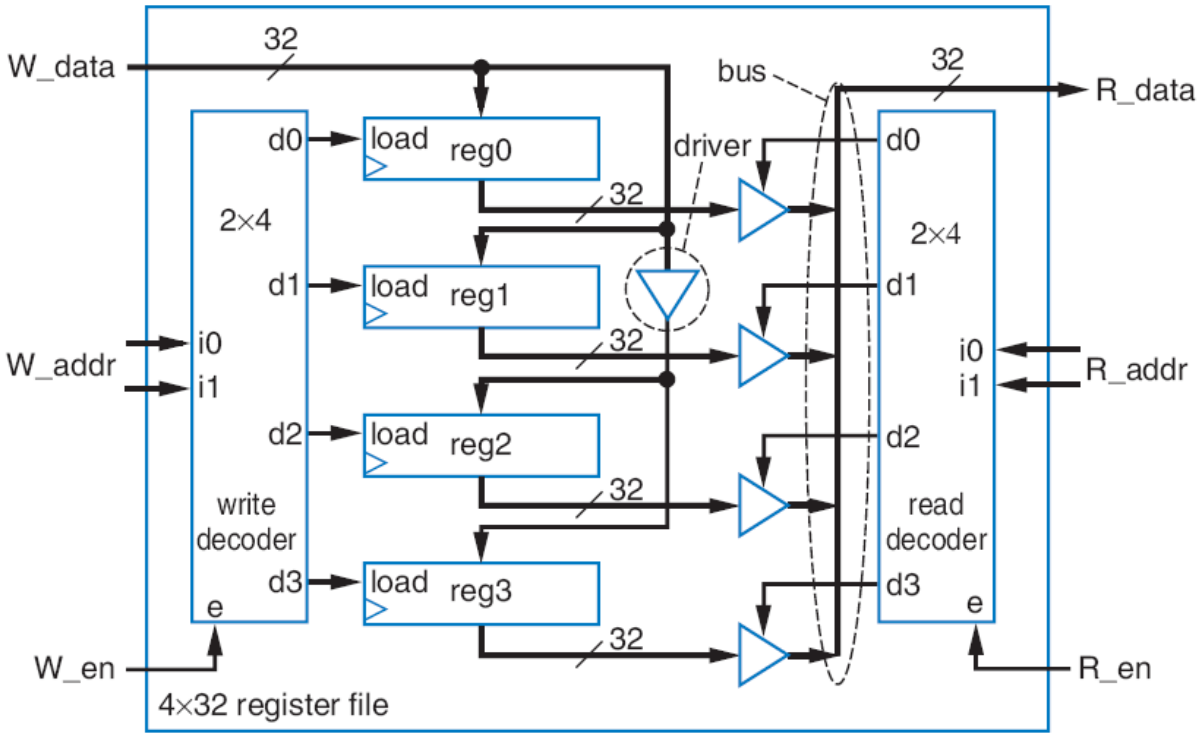
- ***MxN register file*** component provides efficient access to M N-bit-wide registers
  - If we have many registers but only need access one or two at a time, a register file is more efficient
  - Ex: Above-mirror display (earlier example), but this time having 16 32-bit registers
    - Too many wires, and big mux is too slow



Register File

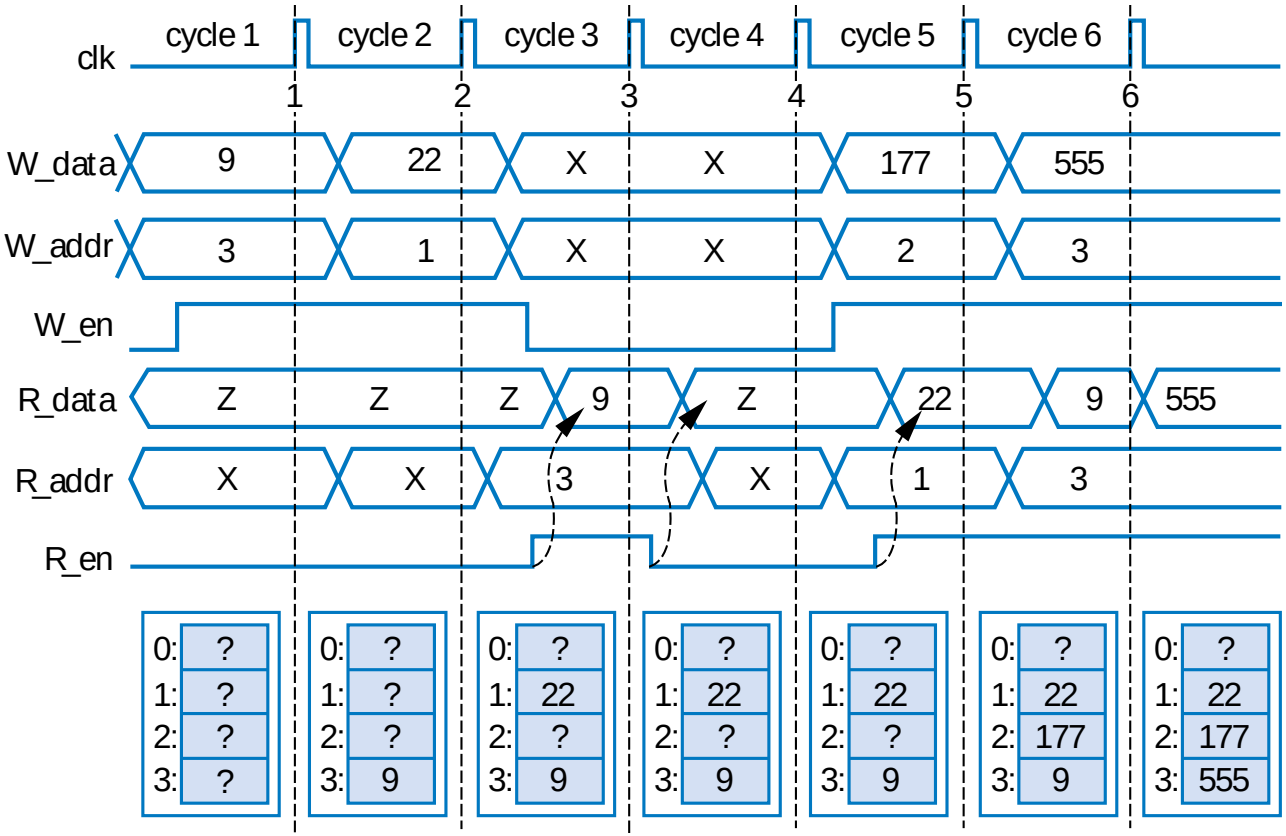
- Instead, want component that has one data input and one data output, and allows us to specify which internal register to write and which to read





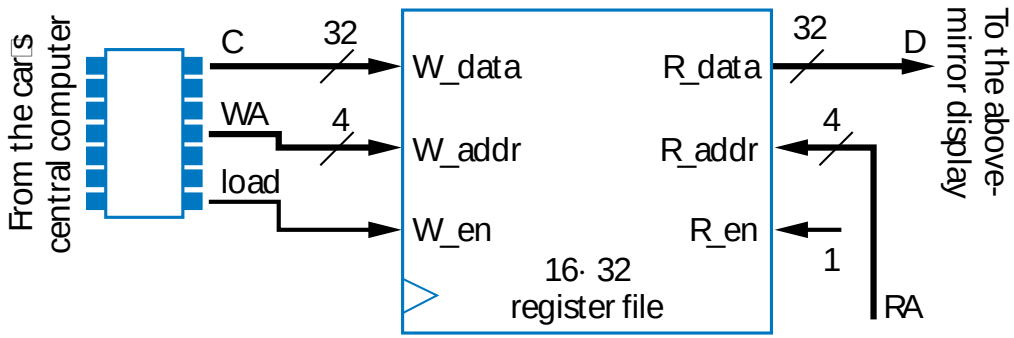
Register File Timing Diagram

- Can write one register and read one register each clock cycle
  - May be same register



Register-File Example: Above-Mirror Display

- 16 32-bit registers that can be written by car's computer, and displayed
  - Use 16x32 register file
  - Simple, elegant design
- Register file hides complexity internally
  - And because only one register needs to be written and/or read at a time, internal design is simple



Chapter Summary

- Need datapath components to store and operate on multibit data
  - Also known as register-transfer-level (RTL) components
- Components introduced
  - Registers
  - Shifters
  - Adders
  - Comparators
  - Counters
  - Multipliers
  - Subtractors
  - Arithmetic-Logic Units
  - Register Files
- Next, we'll combine knowledge of combinational logic design, sequential logic design, and datapath components, to build digital circuits that can perform general and powerful computations